

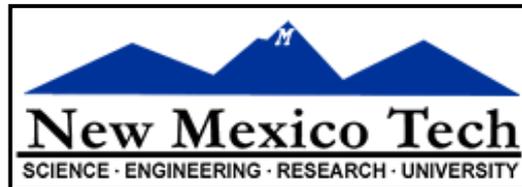
# High Fidelity Imaging

Rick Perley, NRAO-Socorro



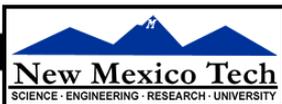
Twelfth Synthesis Imaging Workshop

2010 June 8-15



# 1 What is High Fidelity Imaging?

- Getting the 'Correct Image' – limited only by noise.
  - The best 'dynamic ranges' (brightness contrast) exceed  $10^6$  for some images.
  - (But is the recovered brightness correct?)
- Errors in your image can be caused by many different problems, including (but not limited to):
  - Errors in your data – many origins!
  - Errors in the imaging/deconvolution algorithms used
  - Errors in your methodology
  - Insufficient information
- But before discussing these, and what you can do about them, I show the effect of errors of different types on your image.



## 2 The Effects of Visibility Errors on Image Fidelity

- The most common, and simplest source of error is an error in the measures of the visibility (spatial coherence function).
- Consider a point source of unit flux density ( $S = 1$ ) at the phase center, observed by a telescope array of  $N$  antennas.
- Formally, the sky intensity is:

$$I(l, m) = \delta(l, m)$$

- The correct visibility, for any baseline is:

$$V(u, v) = 1$$

- These are analytic expressions – they presume infinite coverage.
- In fact, we have  $N$  antennas, from which we get, at any one time

$$N_v = \frac{N(N-1)}{2} \text{ visibilities}$$

- Each of these  $N_v$  visibilities is a complex number, and is a function of the baseline coordinates  $(u_k, v_k)$ .

## 2 The Effects of Visibility Errors on Image Fidelity

- The simplest image is made by direction summation over all the visibilities -- (a Direct Transform):

$$I(l, m) = \frac{1}{2N_V} \sum_{k=1}^{N_V} \left[ V_k(u, v) e^{2\pi i(u_k l + v_k m)} + V_k^*(u, v) e^{-2\pi i(u_k l + v_k m)} \right]$$

- For our unit source at the image center, we get

$$I(l, m) = \frac{1}{N_V} \sum_{k=1}^{N_V} \cos \left[ \pi (u_k l + v_k m) \right]$$

- But let us suppose that for one baseline, at one time, there is an error in the amplitude and the phase, so the measured visibility is:

$$V(u, v) = (1 + \varepsilon) \delta(u - u_0, v - v_0) e^{-i\phi}$$

where  $\varepsilon$  = the error in the visibility amplitude

$\phi$  = the error (in radians) in the visibility phase.

## 2 The Effects of Visibility Errors on Image Fidelity

- The map we get from this becomes

$$I(l, m) = \frac{1}{N_V} \left\{ \sum_{k=2}^{N_V} \cos \left[ \pi (u_k l + v_k m) - \phi \right] + (1 + \varepsilon) \cos \left[ \pi (u_1 l + v_1 m) - \phi \right] \right\}$$

- The 'error map' associated with this visibility error is the difference between the image and the 'beam':

$$\Delta I(l, m) = \frac{1}{N_V} \left\{ (1 + \varepsilon) \cos \left[ \pi (u_1 l + v_1 m) - \phi \right] - \cos \left[ \pi (u_1 l + v_1 m) - \phi \right] \right\}$$

- This is a single-(spatial) frequency fringe pattern across the entire map, with a small amplitude and phase offset.
- Let us simplify by considering amplitude and phase errors separately.

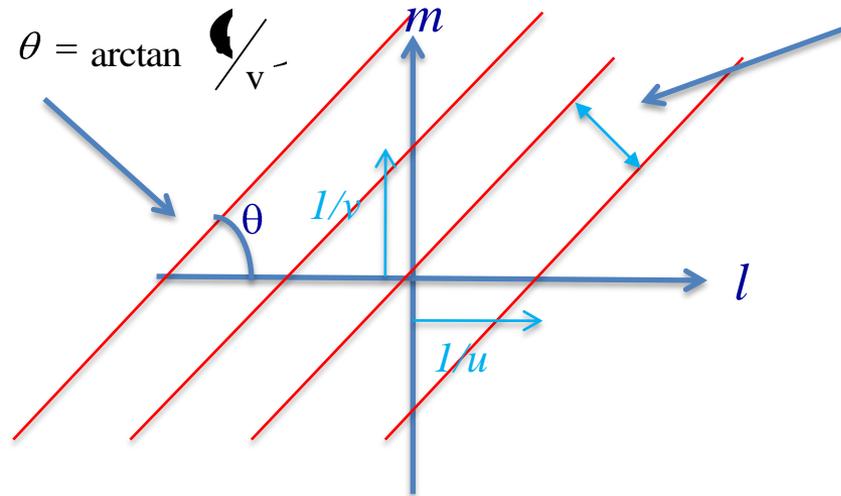
1) Amplitude error only:  $\phi = 0$ . Then,

$$\Delta I = \frac{\varepsilon}{N_V} \cos \left[ \pi (u_1 l + v_1 m) \right]$$

## 2 The Effect of an Amplitude Error on Image Fidelity

$$\Delta I = \frac{\varepsilon}{N_V} \cos \left[ \pi (u_1 l + v_1 m) \right]$$

- This is a sinusoidal wave of amplitude  $\varepsilon/N_V$ , with period  $1/\sqrt{u_1^2 + v_1^2}$  tilted at an angle  $\theta = \arctan \left( \frac{v_1}{u_1} \right)$

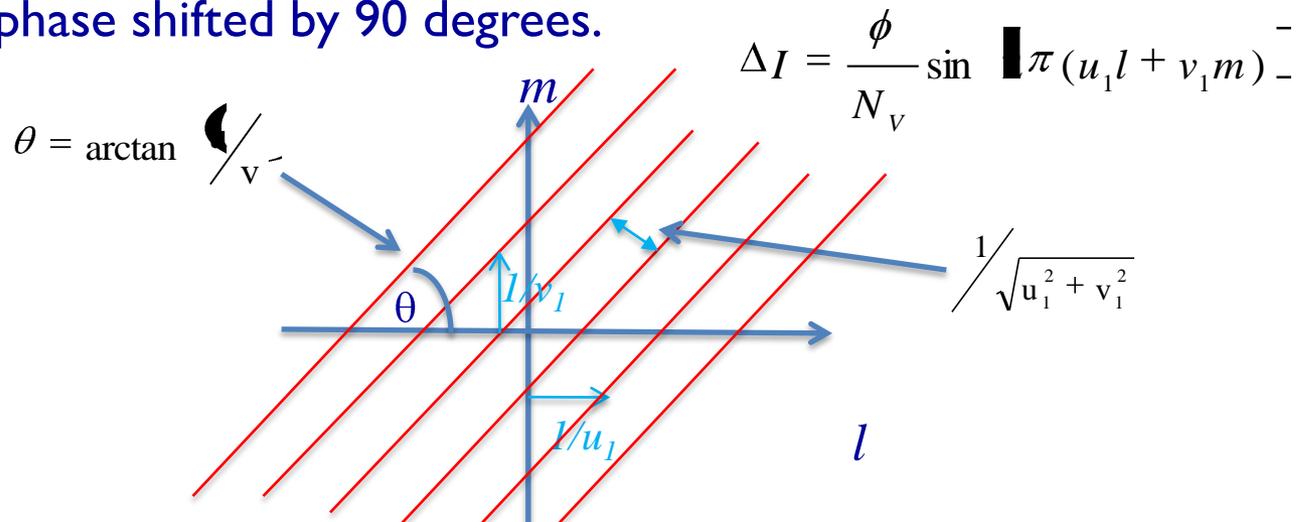


- As an example, if the amplitude error is 10% ( $\varepsilon = 0.1$ ), and  $N_V = 10^6$ , the  $\Delta I = 10^{-7}$  – a very small value!
- Note: The error pattern is even about the location of the source.

## 2 The Effect of a Phase Error on Image Fidelity

In this case: 
$$\Delta I = \frac{1}{N_v} \cos \left[ \pi (u_1 l + v_1 m) - \phi \right] - \cos \left[ \pi (u_1 l + v_1 m) \right]$$

- For small phase error,  $\phi \ll 1$ , 
$$\Delta I = \frac{\phi}{N_v} \sin \left[ \pi (u_1 l + v_1 m) \right]$$
- This gives the same error pattern, but with the amplitude  $\varepsilon$  replaced by  $\phi$ , and the phase shifted by 90 degrees.



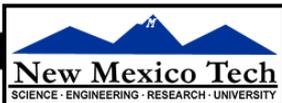
- From this, we derive an **Important Rule**:

A phase error of x radians has the same effect as an amplitude error of 100 x %

- For example, a phase error of 1/10 rad  $\sim$  6 degrees has the same effect as an amplitude error of 10%.

# Amplitude vs Phase Errors.

- This little rule explains why phase errors are deemed to be so much more important than amplitude errors.
- Modern interferometers, and cm-wave atmospheric transmission, are so good that fluctuations in the amplitudes of more than a few percent are very rare.
- But phase errors – primarily due to the atmosphere, but also from the electronics, are always worse than 10 degrees – often worse than 1 radian!
- Phase errors – because they are large – are nearly always the initial limiting cause of poor imaging.



# Errors and Dynamic Range (or Fidelity):

- I now define the dynamic range as the ratio:  $F = \text{Peak}/\text{RMS}$ .
- For our examples, the peak is always 1.0, so the fidelity  $F$  is:

$$F = \frac{\sqrt{2} N_V}{\varepsilon}$$

- For amplitude error of  $100\varepsilon$  %

$$F = \frac{\sqrt{2} N_V}{\phi}$$

- For phase error of  $\phi$  radians

- So, taking our canonical example of 0.1 rad error on one baseline for one single visibility, (or 10% amplitude error):
- $F = 3 \times 10^6$  for  $N_V = 250,000$  (typical for an entire day)
- $F = 5000$  for  $N_V = 351$  (a single snapshot).
- Errors rarely come on single baselines for a single time. We move on to more practical examples now.

# Other Examples of Fidelity Loss

- **Example A: All baselines have an error of  $\sim \phi$  rad at one time.**

Since each baseline's visibility is gridded in a different place, the errors from each can be considered random in the image plane. Hence, the rms adds in quadrature. The fidelity declines by a factor  $\sqrt{N_v} \sim \frac{N}{\sqrt{2}}$

- Thus:  $F = \frac{N}{\varepsilon}$  (N = # of antennas)

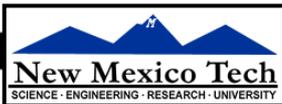
- So, in a 'snapshot',  $F \sim 270$ .

- **Example B: One antenna has phase error  $\varepsilon$ , at one time.**

Here, (N-1) baselines have a phase error – but since each is gridded in a separate place, the errors again add in quadrature. The fidelity is lowered from the single-baseline error by a factor  $\sqrt{N-1}$ , giving

$$F = \frac{N^{3/2}}{\sqrt{2}\varepsilon}$$

- So, for our canonical 'snapshot' example,  $F \sim 1000$



# The Effect of Steady Errors

- **Example C: One baseline has an error of  $\sim \phi$  rad at all times.**

This case is importantly different, in that the error is not randomly distributed in the (u,v) plane, but rather follows an elliptical locus.

- To simplify, imagine the observation is at the north pole. Then the locus of the bad baseline is a circle, of radius  $q = \sqrt{u^2 + v^2}$

- One can show (see EVLA Memo 49 for details) that the error pattern is:

$$\Delta I = \frac{2\varepsilon}{N(N-1)} J_0 \left[ \pi q \theta \right]$$

- The error pattern consists of rings centered on the source ('bull's eye').

- For large  $q\theta$  (this is the number of rings away from the center), the fidelity can be shown to be

$$F = \frac{N(N-1)\pi \sqrt{q\theta}}{\sqrt{2\varepsilon}}$$

- So, again taking  $\varepsilon = 0.1$ , and  $q\theta = 100$ ,  $F \sim 1.6 \times 10^5$ .

# One More Example of Fidelity Loss

- **Example D: All baselines have a steady error of  $\sim \varepsilon$  at all times.**

Following the same methods as before, the fidelity will be lowered by the square root of the number of baselines.

Hence,

$$F = \frac{N(N-1)\pi\sqrt{q\theta}}{\sqrt{2\varepsilon}} \sqrt{\frac{2}{N(N-1)}} \sim \frac{\pi N\sqrt{q\theta}}{\varepsilon}$$

- So, again taking  $\varepsilon = 0.1$ , and  $q\theta = 100$ ,  $F \sim 8000$ .

# Time-Variable Errors

- In real life, the atmosphere and/or electronics introduces phase or amplitude variations. What is the effect of these?
- Suppose the phase on each antenna changes by  $\phi$  radians on a typical timescale of  $\Delta t$  hours.
- Over an observation of  $T$  hours, we can imagine the image comprising  $N_S = \Delta t/T$  individual 'snapshots', each with an independent set of errors.
- The dynamic range on each snapshot is given by

$$F \sim \frac{N}{\varepsilon}$$

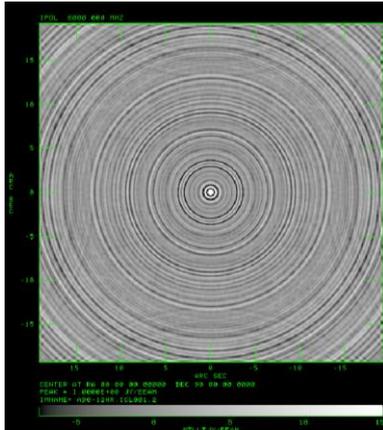
- So, for the entire observation, we get

$$F = \frac{N \sqrt{N_S}}{\varepsilon}$$

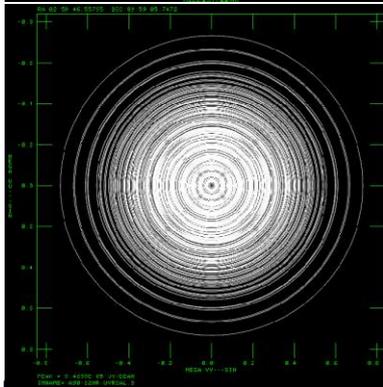
- The value of  $N_S$  can vary from  $\sim 100$  to many thousands.

# Some Examples: Ideal Data

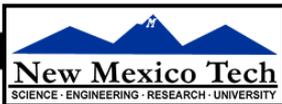
- I illustrate these ideas with some simple simulations.
- EVLA ,  $\nu_0 = 6$  GHz, BW = 4 GHz,  $\delta = 90$ , 'A'-configuration
- Used the AIPS program 'UVCON' to generate visibilities, with  $S = 1$  Jy.



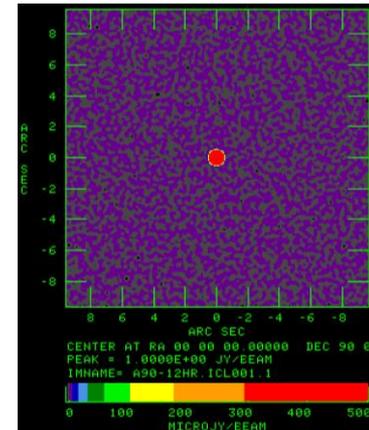
The 'Dirty' Map  
after a 12 hour  
observation.  
Note the 'reflected'  
grating rings.



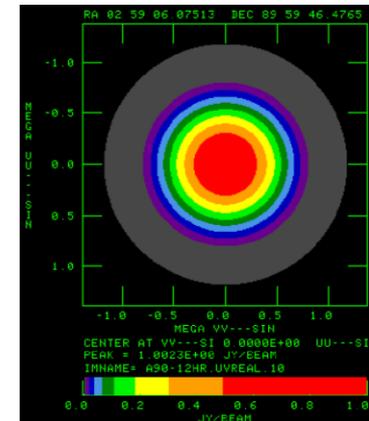
The U-V Coverage  
after a 12 hour  
observation.  
Variations are due  
to gridding.



The 'Clean' Map  
 $1\sigma = 1.3 \mu\text{Jy}$   
Pk = 1 Jy  
No artifacts!



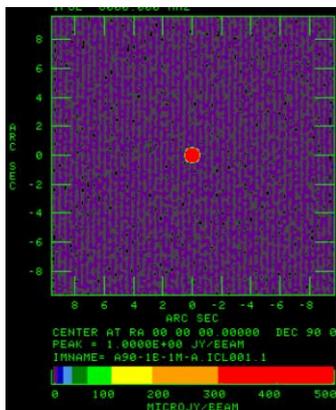
The FT of the 'Clean'  
map  
Note that the  
amplitudes do \*not\*  
match the data!  
The taper comes from  
the Clean Beam.



# One-Baseline Errors – Amplitude Error of 10%

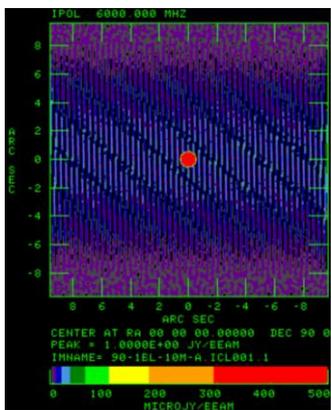
- Examples with a single errant baseline for 1m, 10 m, 1 h, and 12 hours
- $N_v \sim 250,000$

1 minute



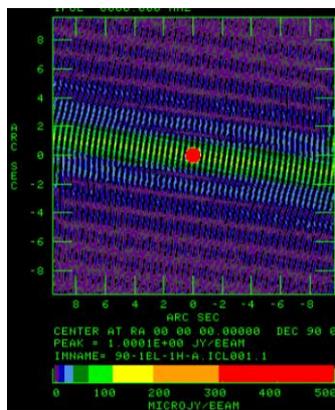
$1 \sigma = 1.9 \mu\text{Jy}$

10 minutes



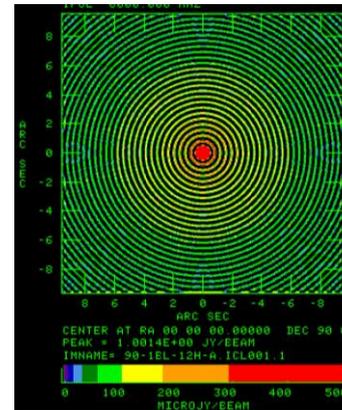
$1 \sigma = 9.4 \mu\text{Jy}$

1 hour



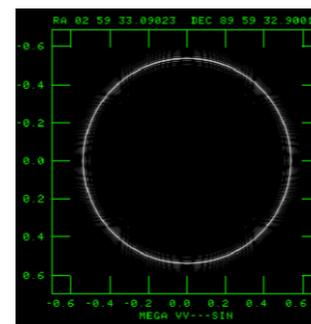
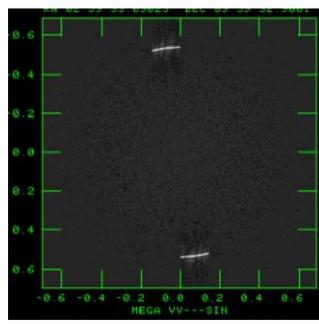
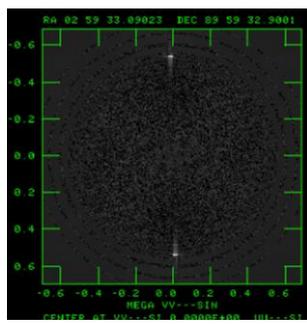
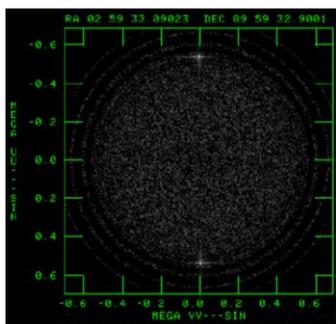
$1 \sigma = 25 \mu\text{Jy}$

12 hours

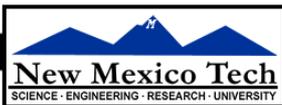


$1 \sigma = 79 \mu\text{Jy}$

The four ‘cleaned’ images, each with peak = 1 Jy. All images use the same transfer function.



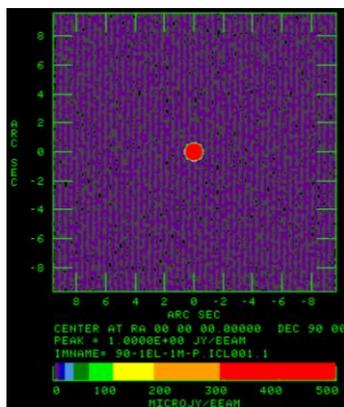
The four U-V plane amplitudes. Note the easy identification of the errors.



# One-Baseline Errors – Phase Error of 0.1 rad = 6 deg.

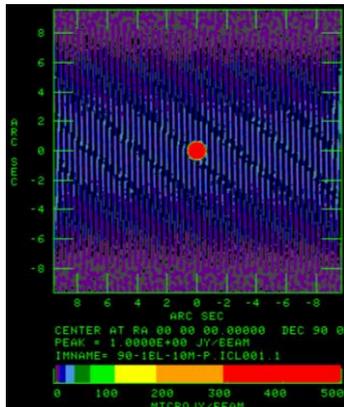
- Examples with a single errant baseline for 1m, 10 m, 1 h, and 12 hours
- $N_v \sim 250,000$

1 minute



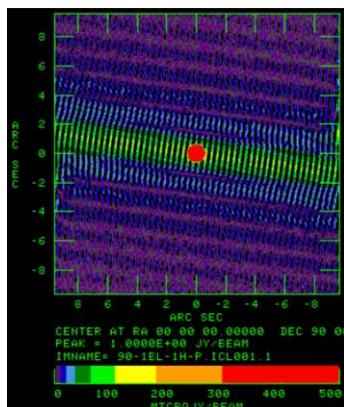
$1 \sigma = 2.0 \mu\text{Jy}$

10 minutes



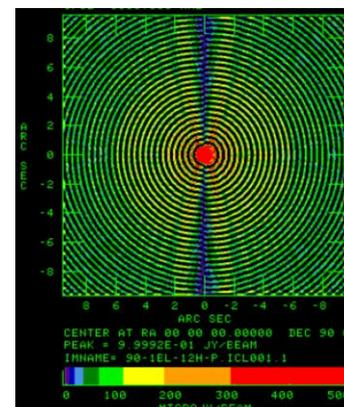
$1 \sigma = 9.8 \mu\text{Jy}$

1 hour



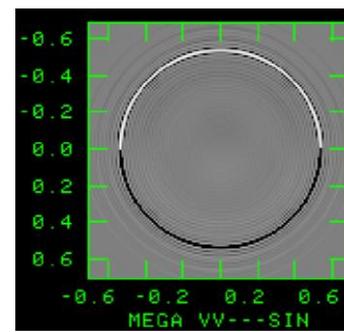
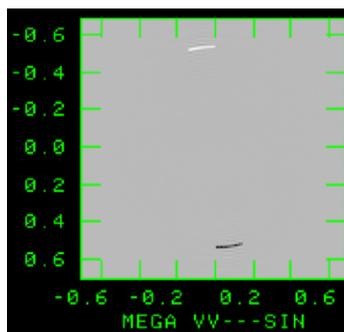
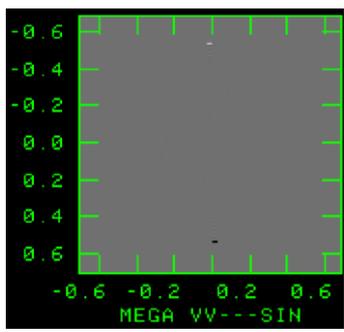
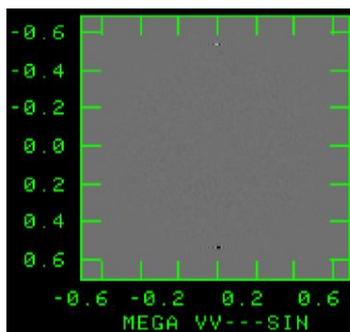
$1 \sigma = 26 \mu\text{Jy}$

12 hours

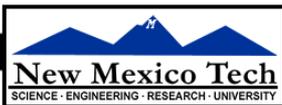


$1 \sigma = 82 \mu\text{Jy}$

The four ‘cleaned’ images, each with peak = 1 Jy. All images use the same transfer function.



The four U-V plane phases. Note the easy identification of the errors.



# One-Antenna Errors – Amplitude Error of 10%

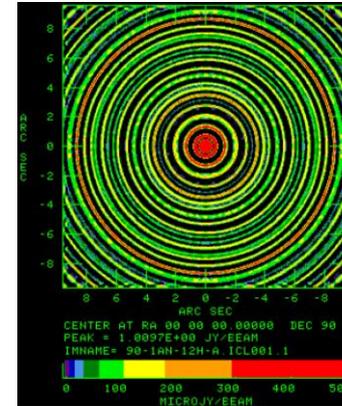
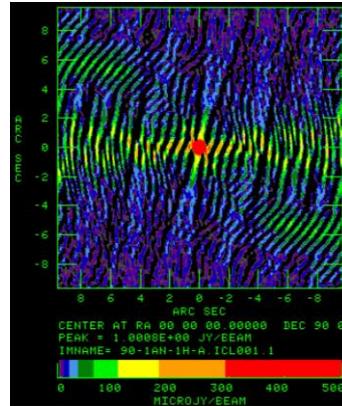
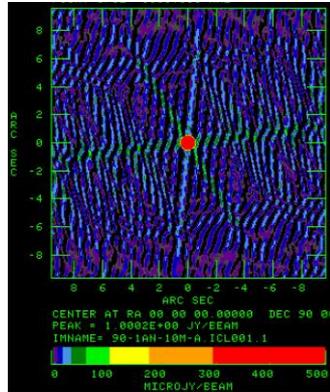
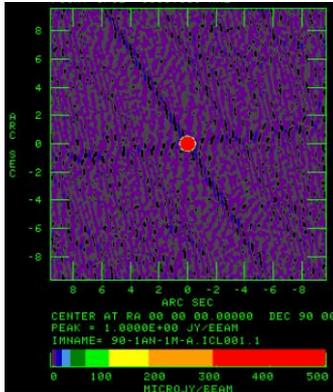
- Examples with a single errant antenna for 1m, 10 m, 1 h, and 12 hours
- $N_v \sim 250,000$

1 minute

10 minutes

1 hour

12 hours



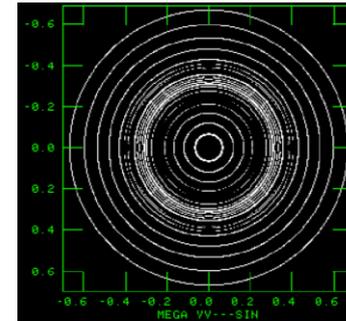
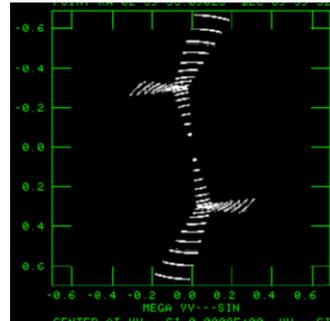
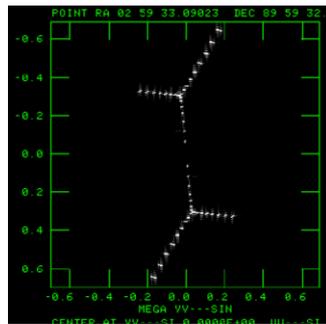
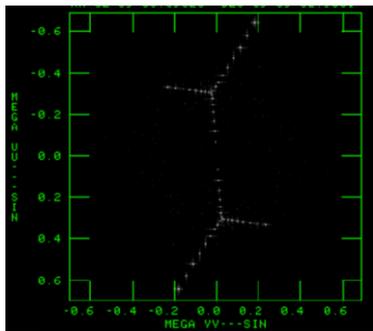
The four ‘cleaned’ images, each with peak = 1 Jy. All images use the same transfer function.

$1 \sigma = 2.3 \mu\text{Jy}$

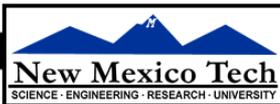
$1 \sigma = 16 \mu\text{Jy}$

$1 \sigma = 42 \mu\text{Jy}$

$1 \sigma = 142 \mu\text{Jy}$



The four U-V plane amplitudes. Note the easy identification of the errors.



# One-Antenna Errors – Phase Error of 0.1 rad = 6 deg.

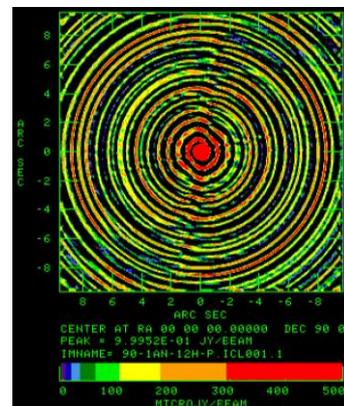
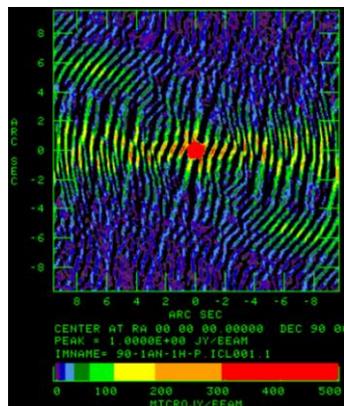
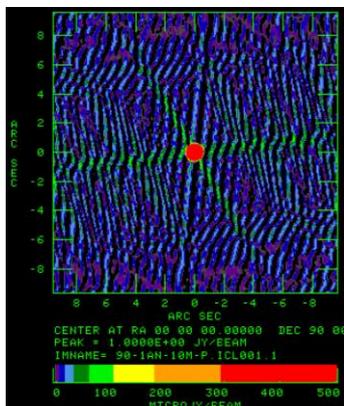
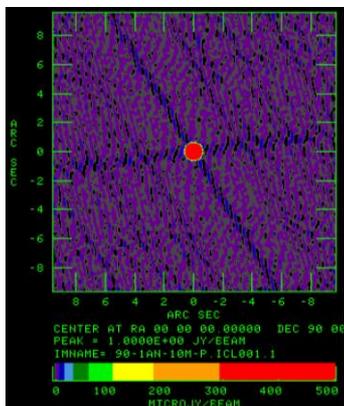
- Examples with a single errant antenna for 1m, 10 m, 1 h, and 12 hours
- $N_v \sim 250,000$

1 minute

10 minutes

1 hour

12 hours



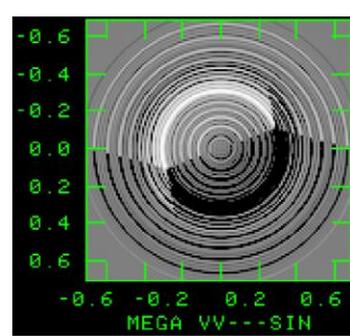
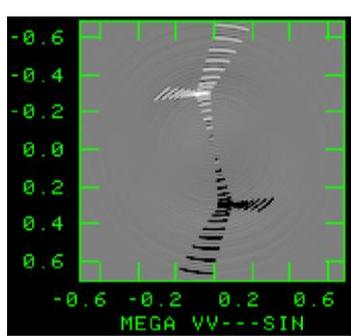
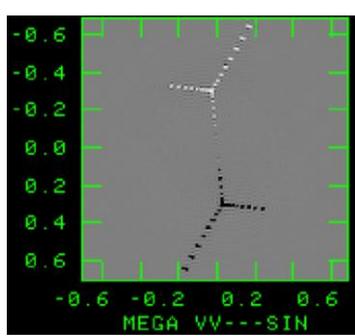
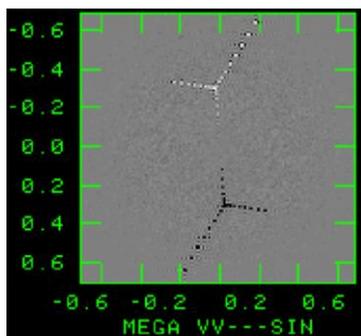
The four ‘cleaned’ images, each with peak = 1 Jy. All images use the same transfer function.

$1 \sigma = 2.9 \mu\text{Jy}$

$1 \sigma = 20 \mu\text{Jy}$

$1 \sigma = 52 \mu\text{Jy}$

$1 \sigma = 147 \mu\text{Jy}$



The four U-V plane phases. Note the easy identification of the errors.

# Finding and Correcting, or Removing Bad Data

- How to find and fix bad data?
- We first must consider the types, and origins, of errors.
- We can write, in general:

$$\tilde{V}_{ij}(t) = g_i(t) g_j^*(t) V_{ij} + g_{ij}(t) V_{ij} + C_{ij}(t) + \varepsilon_{ij}(t)$$

- Here,  $\tilde{V}_{ij}(t)$  is the calibrated visibility, and  $V_{ij}(t)$  is the observed visibility.

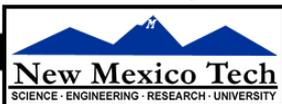
$g_i(t)$  is an antenna based gain

$g_{ij}(t)$  is a multiplicative baseline-based gain.

$C_{ij}(t)$  is an additive baseline-based gain, and

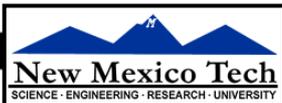
$\varepsilon_{ij}(t)$  is a random additive term, due to noise.

- In principle, the methods of self-calibration are extremely effective at finding and removing all the antenna-based ('closing') errors.
- The method's effectiveness is usually limited by the accuracy of the model.
- In the end, it is usually the 'non-closing' errors which limit fidelity for strong sources.



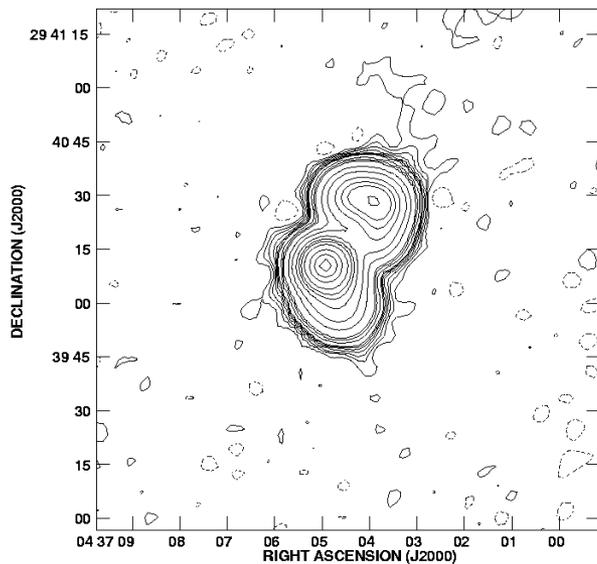
# Finding and Correcting, or Removing Bad Data

- Self-calibration works well for a number of reasons:
  - The most important errors really are antenna-based (notably atmospheric/instrumental phase).
  - The error is 'seen' identically on  $N - 1$  baselines at the same time – improving the SNR by a factor  $\sim \sqrt{N - 1}$ .
  - The  $N - 1$  baselines are of very different lengths and orientations, so the effects of errors in the model are randomized amongst the baselines, improving robustness.
- Non-closing errors can also be calibrated out – but here the process is much less robust! The error is on a single baseline, so not only is the SNR poorer, but there is no tolerance to model errors. The data will be adjusted to precisely match the model you put in!
  - Some (small) safety will be obtained if the non-closing error is constant in time – the solution will then average over the model error, with improved SNR.

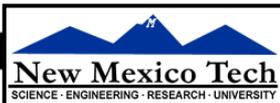
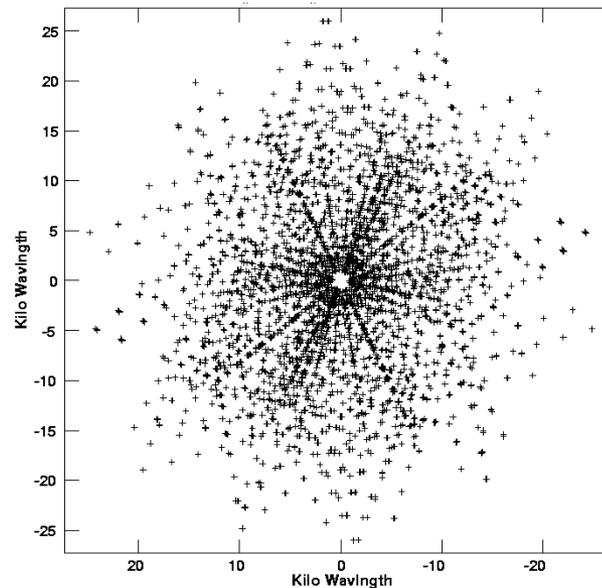


# Finding and Correcting, or Removing Bad Data – a simple example.

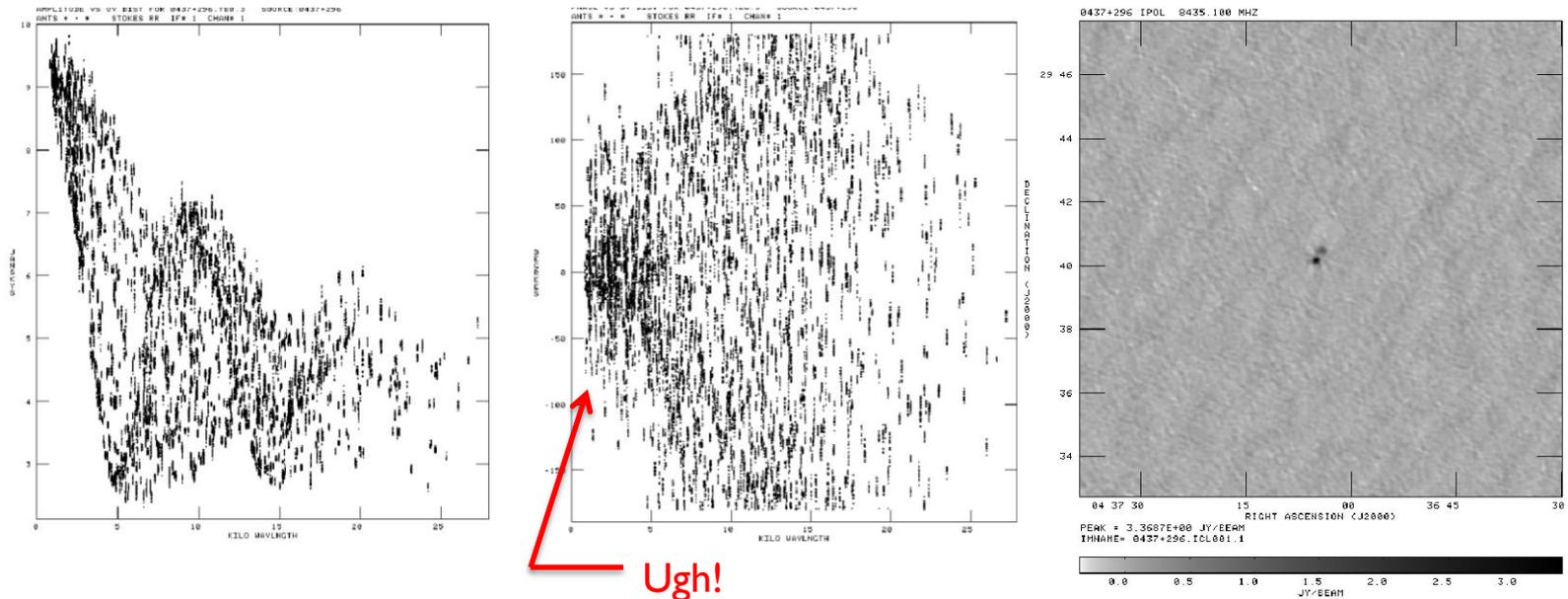
- I show some ‘multiple snapshot’ data on 3C123, a fluxy compact radio source, observed in D-configuration in 2007, at 8.4 GHz.
- There are 7 observations, each of about 30 seconds duration.
- For reference, the ‘best image’, and UV-coverage are shown below.
- Resolution = 8.5 arcseconds. Maximum baseline  $\sim 25 \text{ k}\lambda$



Peak flux = 4.7622E+00 JY/BEAM  
Levs = 4.762E-02 \* (-0.033, 0.033, 0.067, 0.100,  
0.150, 0.200, 0.250, 0.300, 0.500, 0.750, 1, 2.500,  
5, 10, 15, 20, 30, 50, 70, 90)

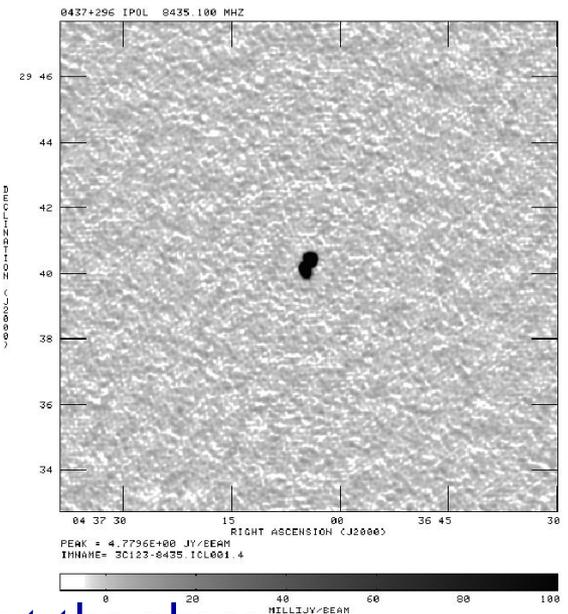
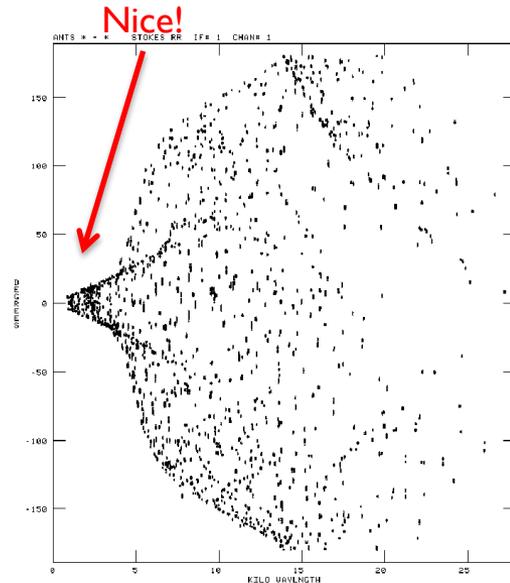
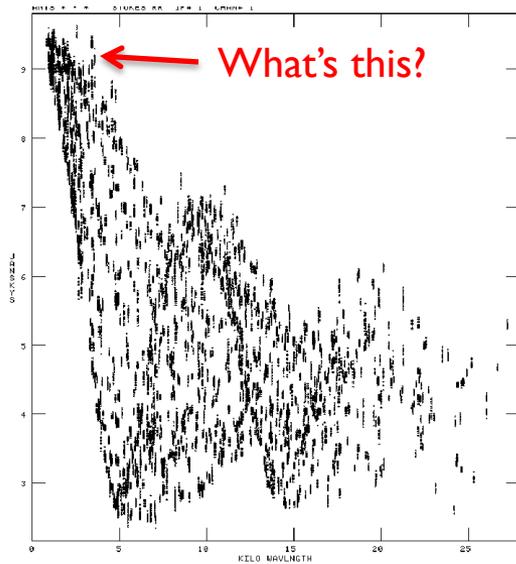


- Following standard calibration against unresolved point sources, and editing the really obviously bad data, the I-d visibility plots look like this, in amplitude and phase:



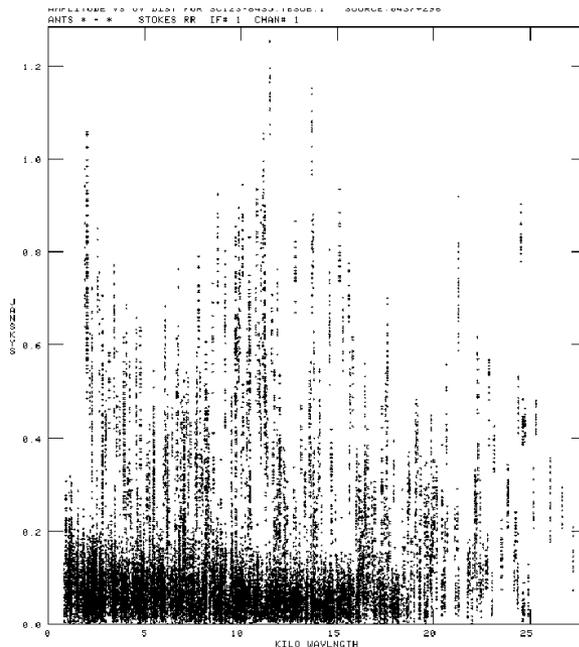
- Note that the amplitudes look quite good, but the phases do not.
- We don't expect a great image.
- Image peak: 3.37 Jy/beam; Image rms = 63 mJy.
- DR = 59 – that's not good!

- Using our good reference image, we do an ‘amplitude and phase’ self-cal.
- The resulting distributions and image are shown below.



- Note that the amplitudes look much the same, but the phase are much better organized..
- Image peak: 4.77 Jy/beam; Image rms = 3.3 mJy.
- DR = 1450 – better, but far from what it should be...

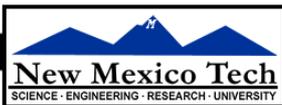
- When self-calibration no longer improves the image, we must look for more exotic errors.
- The next level are ‘closure’, or baseline-based errors.
- The usual step is to subtract the (FT) of your model from the data.
- In AIPS, the program used is ‘UVSUB’.
- Plot the residuals, and decide what to do ...



- If the model matches the data, the residuals should be in the noise – a known value.
- For these data, we expect ~50 mJy.
- Most are close to this, but many are not.

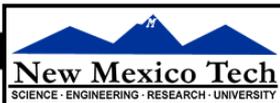
← These are far too large

← These are about right.

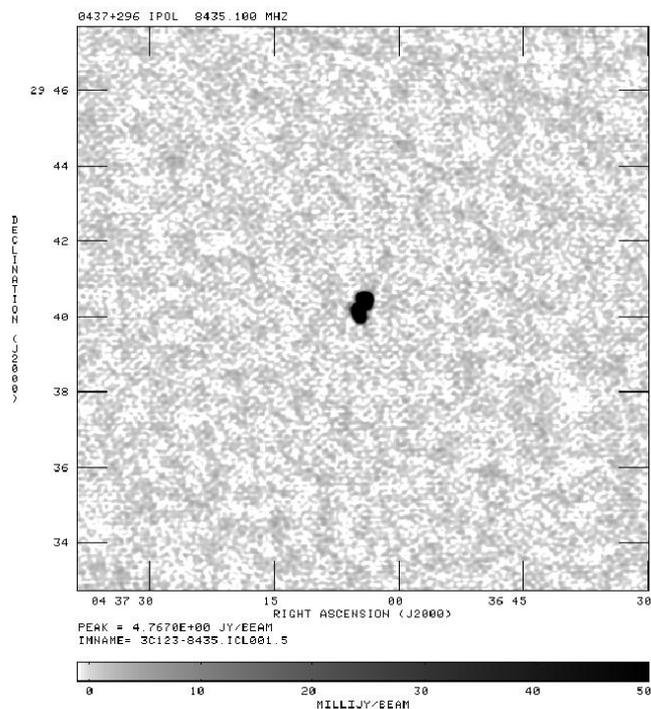


# Removing or Correcting Baseline-based Errors

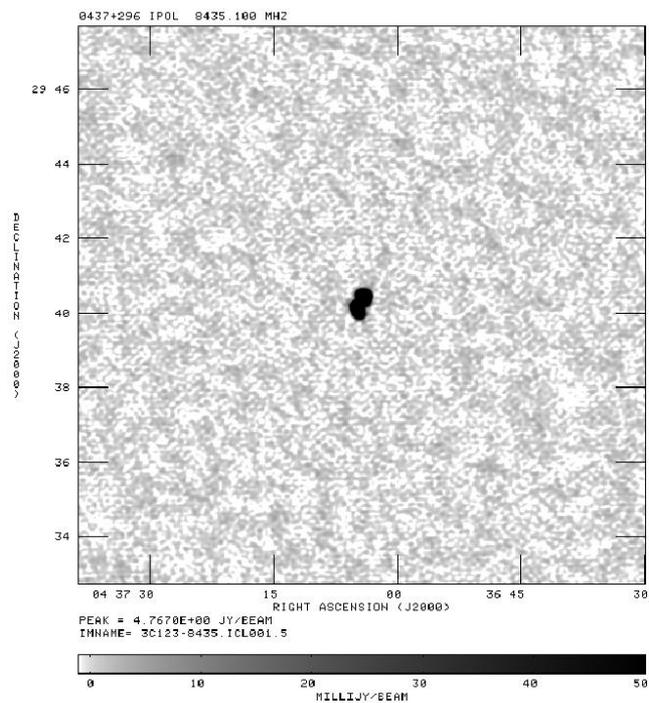
- Once it is determined there are baseline-based errors, the next questions is: What to do about them?
- Solution A: Flag all discrepant visibilities;
- Solution B: Repair them.
- **Solution A:**
  - For our example, I clipped ('CLIP') all residual visibilities above 200 mJy, then restored the model visibilities.
  - Be aware that by using such a crude tool, you will usually be losing some good visibilities, and you will let through some bad ones ...
- **Solution B:**
  - Use the model to determine individual baseline corrections.
  - In AIPS, the program is 'BLCAL'. This produces a set of baseline gains that are applied to the data.
  - This is a powerful – but \*dangerous\* tool ...
  - Since 'closure' errors are usually time invariant, use that condition.



- On Left – Image after clipping high residual visibilities. 20.9 kVis used.
- On Right – Image after correcting for baseline-based errors.



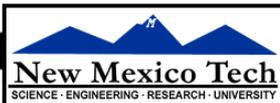
Peak = 4.77 Jy  $s = 1.2$  mJy  
DR = 3980



Peak = 4.76 Jy  $s = 0.83$  mJy  
DR = 5740

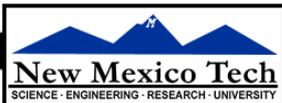
# Law of Diminishing Returns or Knowing When to Quit

- I did not proceed further for this example.
- One can (and many do) continue the process of:
  - Self-calibration (removing antenna-based gains)
  - Imaging/Deconvolution (making your latest model)
  - Visibility subtraction
  - Clipping residuals, or a better baseline calibration.
  - Imaging/Deconvolution
- The process always asymptotes, and you have to give it up, or find a better methodology.
- Note that not all sources of error can be removed by this process.

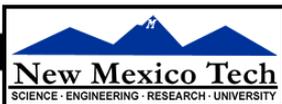


# Sources of Error

- I conclude with a short summary of sources of error.
- This list is necessarily incomplete.
- **Antenna-Based Errors**
  - Electronics gain variations – amplitude and phase – both in time and frequency.
    - Modern systems are very stable – typically 1% in amplitude, a few degrees in phase
  - Atmospheric (Tropospheric/Ionospheric) errors.
    - Attenuation very small at wavelengths longer than  $\sim 2$  cm – except through heavy clouds (like thunderstorms) for 2 – 6 cm.
    - Phase corruptions can be very large – tens to hundreds of degrees.
    - Ionosphere phase errors dominate for  $\lambda > 20$ cm.
  - Antenna pointing errors: primarily amplitude, but also phase.
  - Non-isoplanatic phase screens



- **Baseline-Based Errors** – this list is much longer
  - System Electronics.
    - Offsets in a particular correlator (additive)
    - Gain (normalization?) errors in correlator (multiplicative)
    - Other correlator-based issues (WIDAR has ‘wobbles’ ...)
    - Phase offsets between COS and SIN correlators
    - Non-identical bandpasses, on frequency scales smaller than channel resolution.
    - Delay errors, not compensated by proper delay calibration.
    - Temporal phase winds, not resolved in time (averaging time too long).

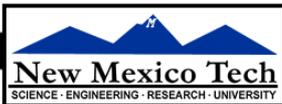


- Impure System Polarization  $V_r' = V_r + D_r V_l$ 
  - Even after the best regular calibration, the visibilities contain contaminants from the other polarizations
  - For example, for Stokes 'I', we can write:

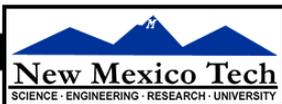
$$V_I' = V_I + D_1 V_Q + D_2 V_U + D_3 V_V$$

- The 'I' visibility has been contaminated by contributions from Q, U, and V, coupled through by complex 'D' factors which describe the leakage of one polarization into the other.
- This term can be significant – polarization can be 30% or higher, and the 'D' terms can be 5%
- The additional terms can easily exceed 1% of the Stokes 'I'.
- Polarization calibration necessary – but note that the antenna beam is variably polarized as a function of angle.

- **Other, far-out effects (to keep you awake at night ...)**
- Correlator quantization correction
  - Digital correlators are non-linear – they err in the calculation of the correlation of very strong sources.
  - This error is completely eliminated with WIDAR.
- Non-coplanar baselines.
  - Important when  $\frac{\lambda_B}{D^2} \geq 1$
  - Software exists to correct this.
- Baseline errors: incorrect baselines leads to incorrect images.
  - Apply baseline corrections to visibility data, perhaps determined after observations are completed.
- Deconvolution Algorithm errors
  - CLEAN, VTESS, etc. do not \*guarantee\* a correct result!
  - Errors in data, holes in the coverage, absence of long or short spacings will result in incorrect images.
  - Best solution – more data!



- Wide-band data
  - New instruments (like EVLA) have huge fractional bandwidths
  - Image structure changes dramatically
  - Antenna primary beams change dramatically
  - New algorithms are being developed to manage this.
- Distant structure
  - In general, antennas ‘sense’ the entire sky – even if the distant structure is highly attenuated. (This problem is especially bad at low frequencies ...)
  - You are likely interested in only a part of the sky.
  - You probably can’t afford to image the entire hemisphere ...
  - Some form of full-sky imaging will be needed to remove distant, unrelated visibilities.
  - Algorithms under development for this.



# How Good Can It Get?

- Shown is our best image (so far) from the EVLA.
- 3C147, with 'WIDAR0' – 12 antennas and two spectral windows at L-band (20cm).
- Time averaging 1 sec.
- BW averaging 1 MHz
- BW 2 x 100 MHz
- Peak = 21200 mJy
- 2<sup>nd</sup> brightest source 32 mJy
- Rms in corner: 32  $\mu$ Jy
- Peak in sidelobe: 13 mJy – largest sidelobes are around this!
- DR ~ 850,000!
- Fidelity quite a bit less.

