

System-level interfaces and performance evaluation methodology for 5G physical layer based on non-orthogonal waveforms

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■ What is 5GNOW?

5GNOW (5th Generation Non-Orthogonal Waveforms for Asynchronous Signalling) is an European collaborative research project supported by the European Commission within FP7 ICT Call 8.

■ Who is in the consortium?

- Fraunhofer HHI (coordinator), Germany, *Dr. Gerhard Wunder*
- Alcatel Lucent (technical coord.), Germany, *Dr. Stephan ten Brink*
- Technische Universität Dresden, Germany, *Prof. Gerhard Fettweis*
- CEA-LETI, France, *Dimitri Ktenas*
- IS-Wireless, Poland, *Dr. Sławomir Pietrzyk*
- National Instruments, Hungary, *Dr. Bertalan Eged*



Vision: 5GNOW is the physical layer evolution of mobile communication network technology such as LTE-Advanced towards emerging application challenges.

- (1) 5G NOW: Why new waveforms?**
- (2) PHY abstraction fundamentals & challenges**
- (3) Proposed Link-to-System interface for FBMC**
- (4) System Simulations: Comparison of FBMC and LTE/OFDMA**

(1) 5G NOW: Why new waveforms?

(2) PHY abstraction fundamentals & challenges

(3) Proposed Link-to-System interface for FBMC

**(4) System Simulations: Comparison of FBMC and
LTE/OFDMA**

Why new waveforms?

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- 5G on the way... intriguing applications
 - Internet of Things
 - Tactile Internet
 - Gigabit Wireless
- Particularly: Fragmented spectrum, spectrum agility

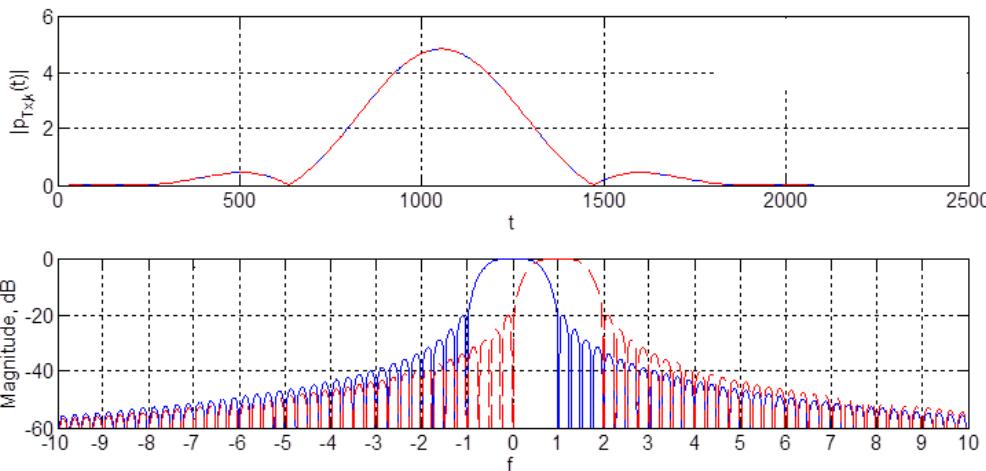
Conclusion: highly flexible robust new waveforms required!

- Several candidate waveforms proposed:
 - Filter Banks Multi-Carrier (**FBMC**)
 - Universal Filtered Multi-Carrier (**UFMC**)
 - Generalized Frequency Division Multiplexing (**GFDM**)
- In this talk we exemplarily focus on FBMC

Why new waveforms?

FBMC Fundamentals

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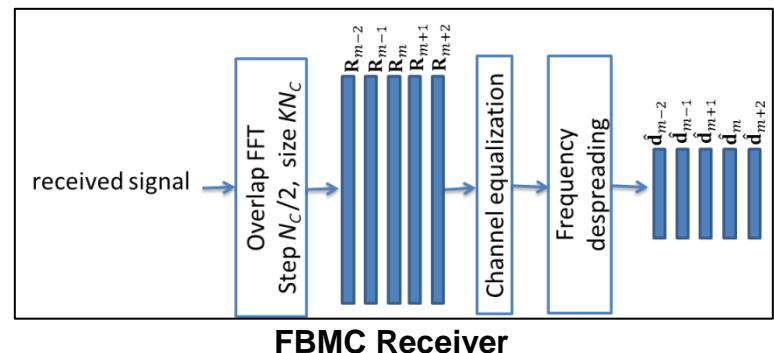
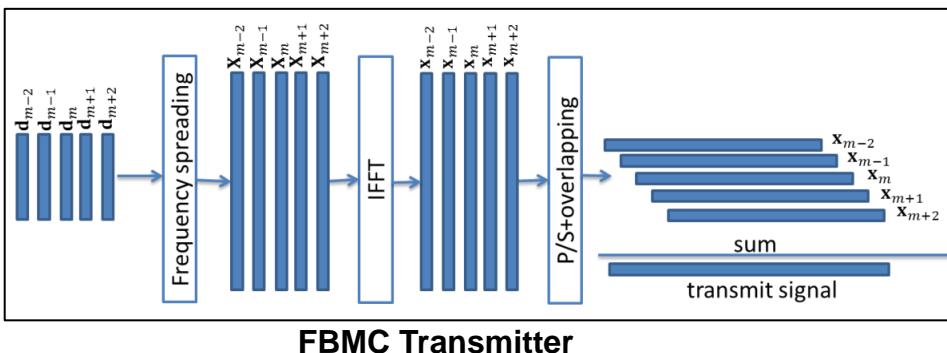
-> Significant Overlap

Features :

- Spectrally well shaped prototype filter
- Overlapped time symbols

Advantages

- No Cyclic Prefix
- Almost perfect separation of frequency subbands without strict synchronization
- Suitability for fragmented spectrum and for CoMP



- FBMC well-established waveform but **system-level evaluations** needed

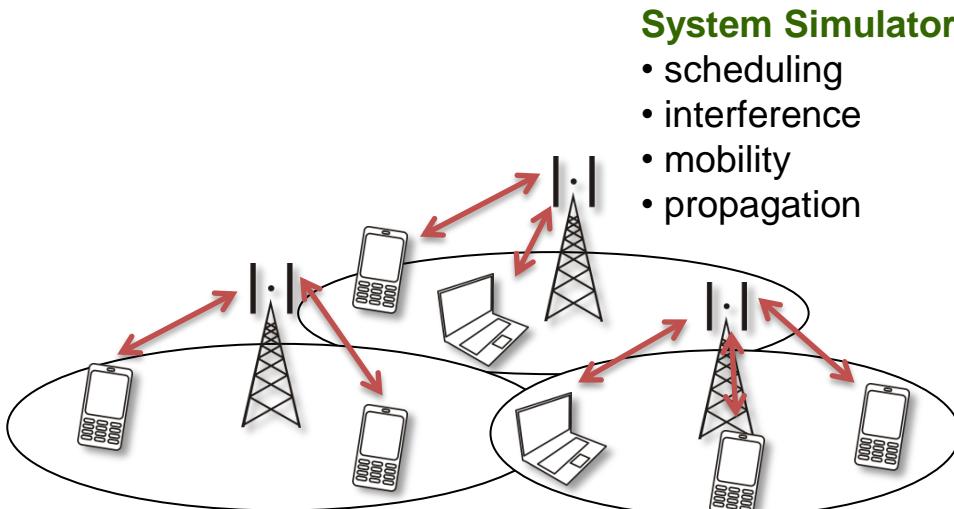
(1) 5G NOW: Why new waveforms?

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(3) Proposed Link-to-System interface for FBMC

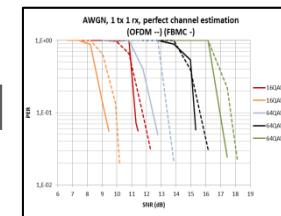
(4) System Simulations: Comparison of FBMC and
LTE/OFDMA

PHY abstraction needed: Obtain *Block Error Rate (BLER)* for a *transport block*, given particular subcarrier channel realizations, *without* having to carry out real signal processing.



System Simulator

- scheduling
- interference
- mobility
- propagation



Link-To-System Interface

Link Level Simulations

- Multiplexing
- Channel Coding, Modulation
- Signal Processing
- Equalization, Detection



- Effective SINR for whole transport block needed
- Challenges:
 - Transport block of bits is spread across (possibly non-orthogonal) subcarriers and each individual subcarrier may experience different SNIR
 - Interferences from different eNBs may have different influence on different subcarriers
 - Each user may use different BW parts and BLER for each may be calculated separately (the UE may use different MCS on allocated resources)

Effective SINR Mapping (ESM) function $\mathcal{J}(\cdot)$ (e.g. Linear, Logarithmic, Exponential, Mutual Information) needed, such that

$$\gamma_{eff} = \mathcal{J}^{-1} \left(\frac{1}{K} \sum_{k=1}^K \mathcal{J}(\gamma_k) \right).$$

- To evaluate system-level performance of FBMC , a PHY abstraction that can deal with *non-orthogonal waveforms* and *synchronization errors* is needed.

Elements to consider:

- (1) Define suitable quality measure per subcarrier
- (2) Derive means to incorporate temporal-spectral asynchronisms
- (3) Find suitable effective SINR mapping for whole transmission
- (4) Define suitable frame structure and generate AWGN - SNR/BLER tables

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PHY Abstraction for FBMC Information measure

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- FBMC induces additional interference from neighboring subcarriers due to non-orthogonality

Use SNIDR model:

$$\text{SNIDR}_k = \frac{P_s}{P_i + P_n + P_d}$$

Residual Distortion

$$P_d(k) = \frac{2P_s}{K^3} \left\| \frac{H'_k(\omega)}{H_k(\omega)} \right\|^2 C_f$$

 A. Oborina, C. Ibars, L. Guipponi, F. Bader, "Link Performance Model for System Level Simulations of Filter Bank Multicarrier-Based Systems in PMR Networks", *Proceedings ISWCS 2013*,

- In Rayleigh fading channel: $\frac{H'_k(\omega)}{H_k(\omega)} = \frac{1}{w_k} - \frac{w_k}{\sigma^2}$, with $w_k = \frac{2\pi}{N}(k-1)$, $k = 1, \dots, N$
- $H_k(\omega)$, $H'_k(\omega)$: channel frequency response and its derivative evaluated at the k 'th subcarrier frequency.
- P_s, P_i, P_n : signal power, interference power, noise power
- C_f : first order derivative of receiving prototype pulse. The smoother the filter the lower C_f and the lower is the distortion power.

PHY Abstraction for FBMC

Incorporate Time-Frequency Offsets

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The following relation holds for non-random time-frequency shifts

$$\text{sinr} \geq \frac{|A_{h\gamma}(d, \nu)|^2}{\frac{1}{SNR} + B_\gamma - |A_{h\gamma}(d, \nu)|^2}$$



P. Jung, G. Wunder "On Time-Variant Distortions in Multicarrier Transmission With Application to Frequency Offsets and Phase Noise", in *IEEE Transactions on Communications*, Vol. 53, No. 9, pp. 1561-1570, September 2005

- Given a **frequency offset** ν in the OFDM case it holds

$$\text{sinr} \geq \frac{\sin^2 \pi \hat{\nu}/(\pi \hat{\nu})^2}{\frac{\sigma^2}{\epsilon \|p_h\|_1} + (1 - \sin^2 \pi \hat{\nu}/(\pi \hat{\nu})^2)} \quad \text{with} \quad \hat{\nu} = \frac{\nu}{F}$$

- In FBMC case we get :

$$\text{sinr} \geq \frac{|A_{hh}(0, \tilde{\nu})|^2}{\frac{1}{SNR} + 1 - |A_{hh}(0, \tilde{\nu})|^2}$$

- FBMC case (only frequency shift): ambiguity function given by

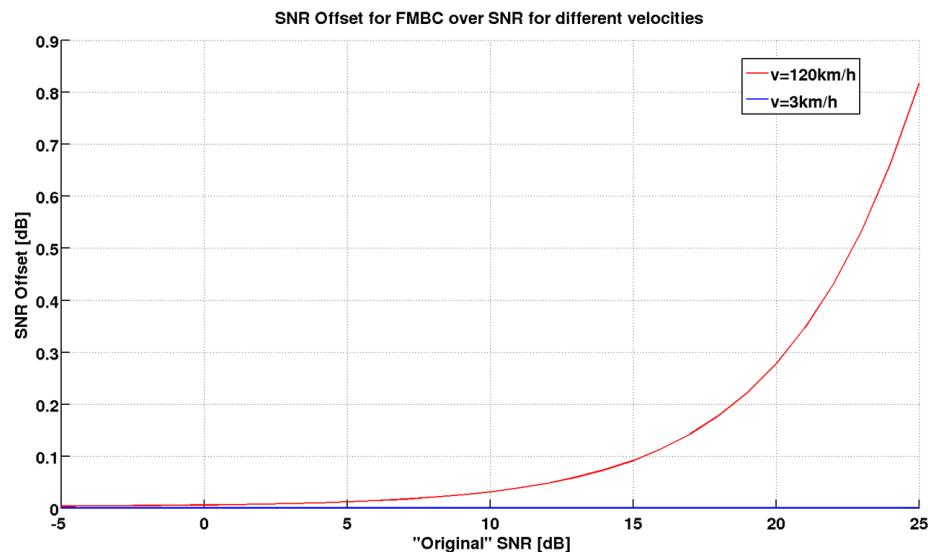
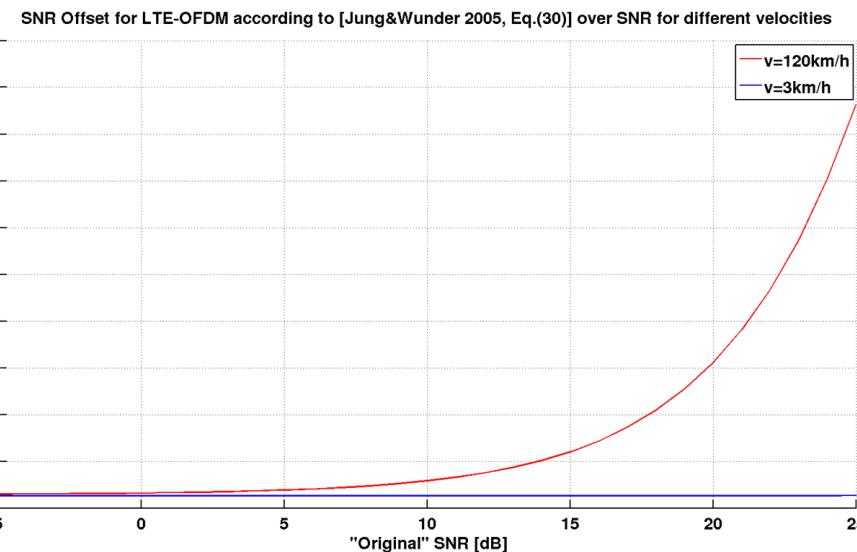
$$A_{hh}(0, \nu) = \int_{-\infty}^{\infty} |\tilde{h}(t)|^2 e^{j2\pi\nu t} dt \quad \text{with} \quad \tilde{h}(t) = \frac{1}{\|h(t)\|_2} h(t).$$

- Transmit (and receive) filter: $h(t) = P_0 + 2 \sum_{k=1}^{K-1} (-1)^k P_k \cos\left(\frac{2\pi k}{KN_c}(t+1)\right), t = 0: KN_c - 2$

PHY Abstraction for FBMC

Frequency offsets due to user velocity

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- Example: Frequency offset induced by user velocity (Doppler shift)
- So far influence of user velocity in system simulator not sufficiently considered

Result:

- Influence of frequency offsets in FBMC much lower
- However small offsets for the considered velocities

Two main challenges:

- (1) Get average quality measure for resource block
 - Use ESM or simple averaging of subcarrier SNIDRs (same as OFDM)
- (2) Get overall effective SNIR for whole transmission: ESM needed
 - Popular method: EESM, but difficult to configure
 - MIESM: Use mutual information as information measure function $\mathcal{I}(\cdot)$

MIESM mapping:

- Good match with AGWN curves without specific parameter optimization.
- A polynomial approximation of mutual information is given by:

$$MI(SINR, m) = \frac{1}{([s \cdot MI_{Shannon}(SINR)]^{-w} + m^{-w})^{\frac{1}{w}}}$$



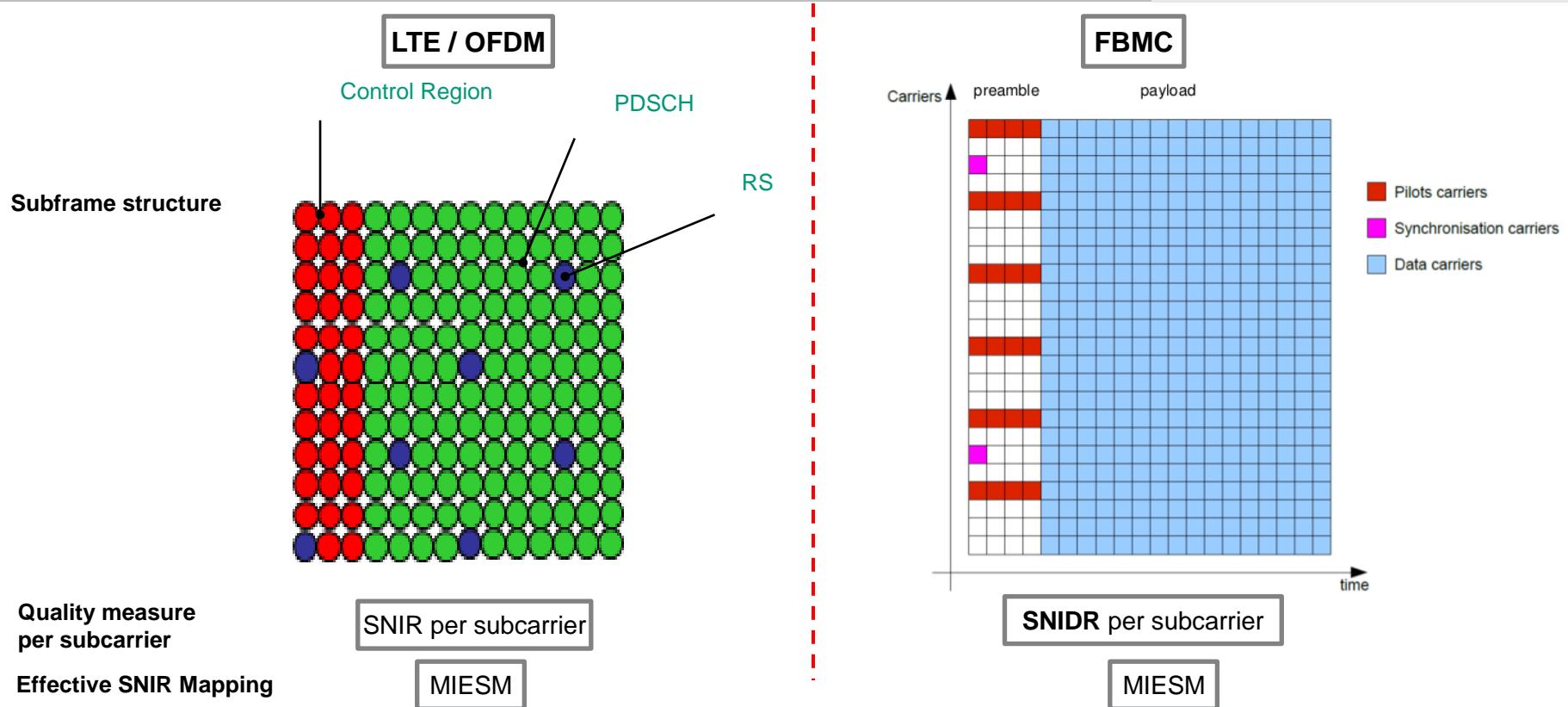
R. Schoenen, C. Teijeiro and D. Bültmann "System Level Performance Evaluation of LTE with MIMO and Relays in Reuse-1 IMT-Advanced Scenarios", in International Conference on Wireless Communications, Networking and Mobile Computing – WiCom, 2010

$$MI_{Shannon}(SINR) = \log_2(1 + 10^{\frac{SINR}{10dB}}), \quad s = s(m) = 0.95 - 0.08 \cdot (m \bmod 2), w = w(m) = 2 \cdot m + 1$$

PHY Abstraction for FBMC

Frame Structure

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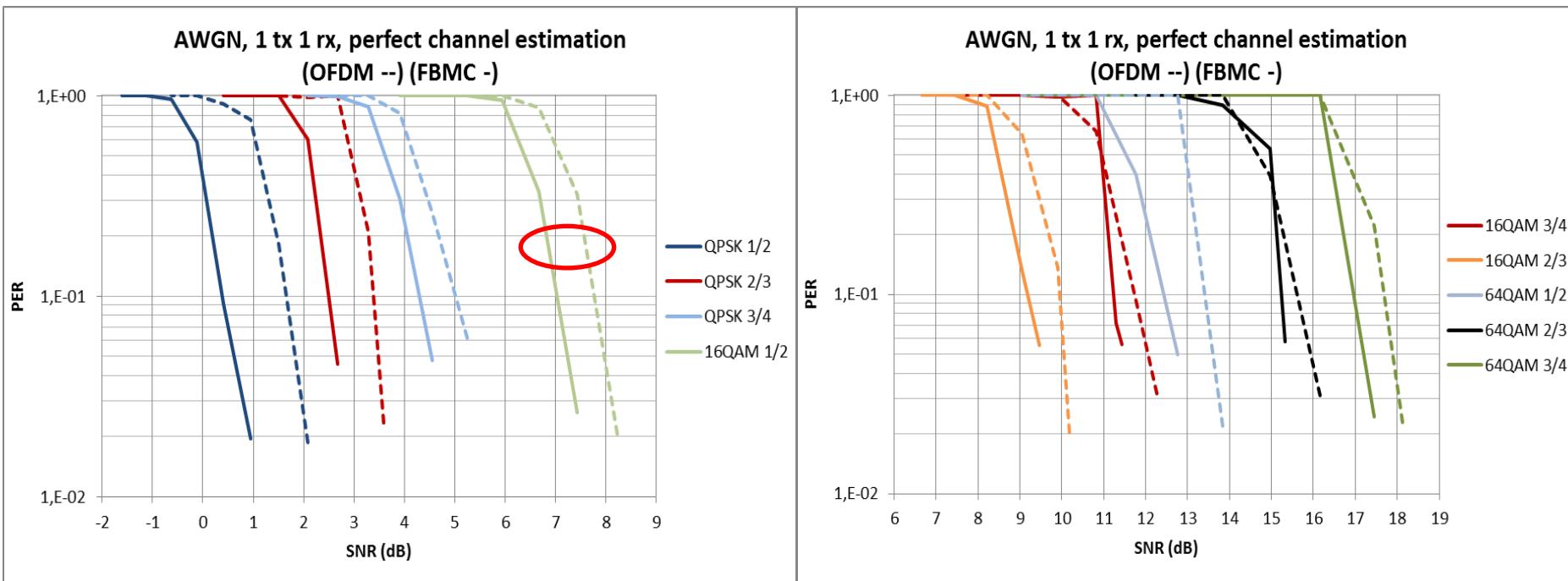


- Frame structure matched to enable fair comparison (to have same throughput)
- OFDM: 9 data and 4 signaling/control/... symbols, subframe duration: 0.93 ms, cyclic prefix: 72 carriers
- FBMC: 16 data and 4 signaling/control/... symbols, subframe duration: 1.63 ms

PHY Abstraction for FBMC Parameters and SNR – BLER curves

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Resulting SNR/BLER curves:



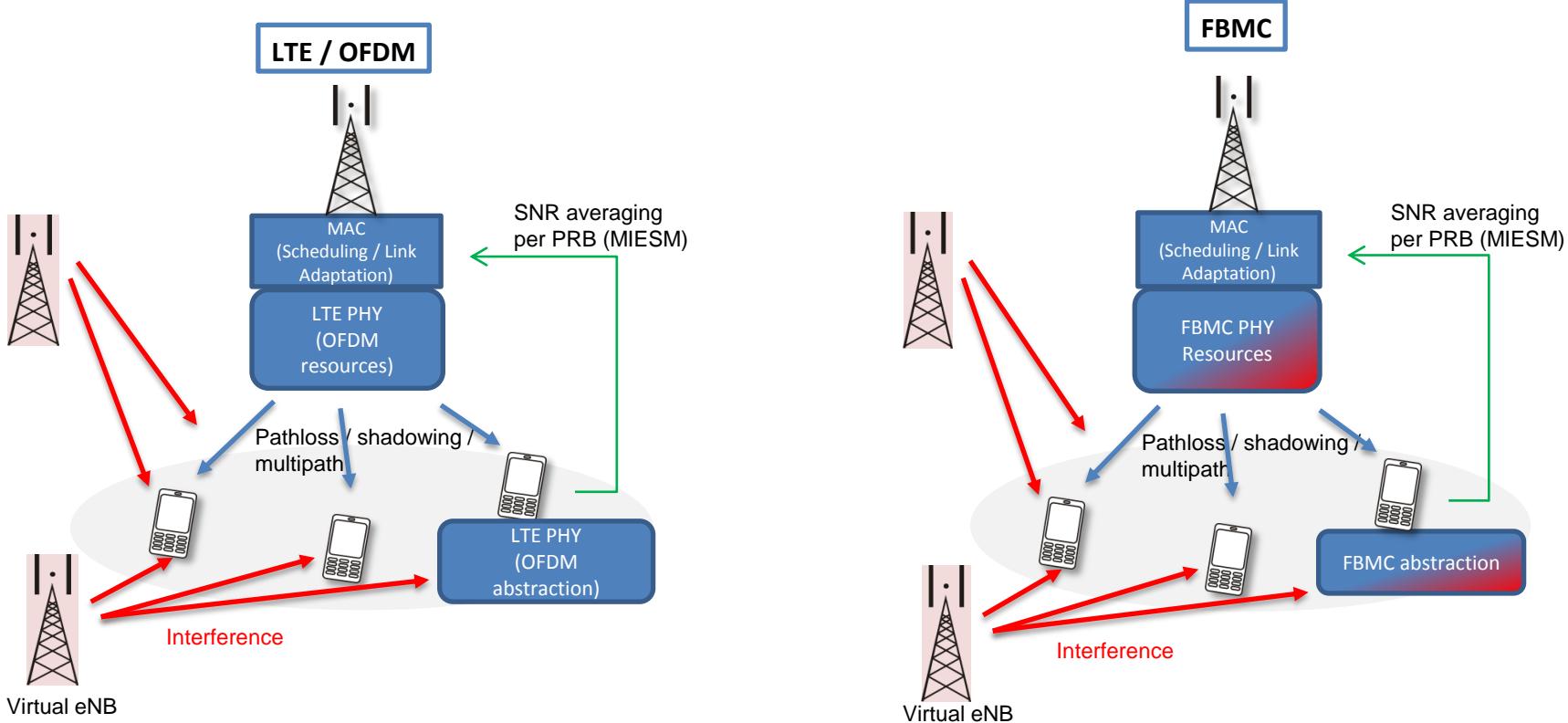
- Convolutional encoder / Viterbi decoder
- Gain through cyclic prefix

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System Simulations

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- System simulations of OFDM vs. FBMC downlink using *LTE MAC Lab* software
 - Number of users: 50, user velocities: 3km/h and 120 km/h, Max-SINR scheduler
 - Duration: 2000/3260 TTIs for FBMC/OFDM, average 6 simulation runs
 - System bandwidth: 10MHz, System band: 2GHz, Ideal feedback

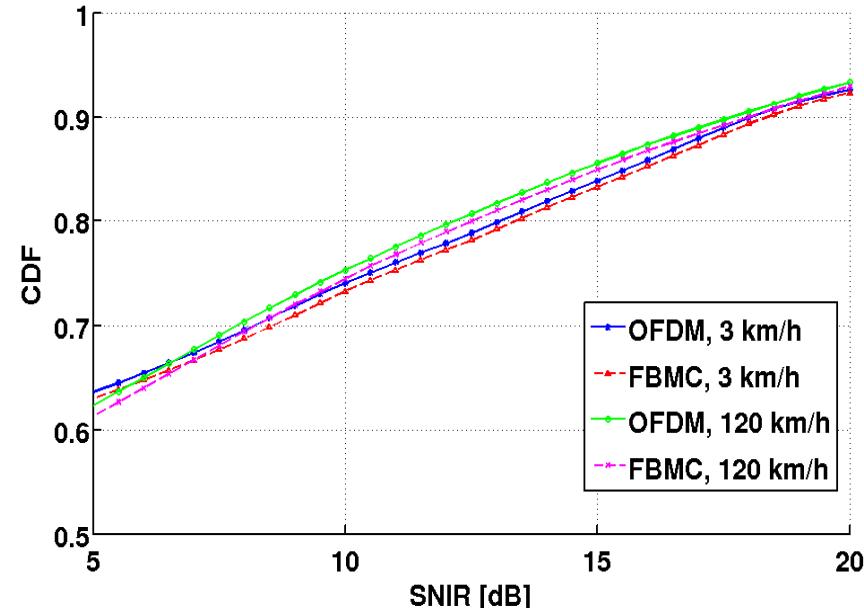


Simulation Results

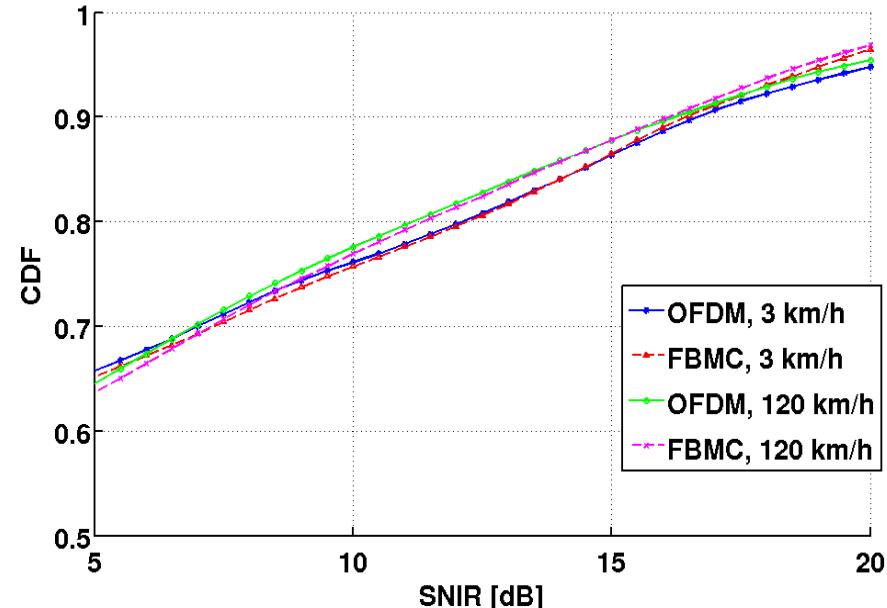
SNIR comparison

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SNIR comparison of OFDM and FBMC. Coherence bandwidth = 398,41 kHz



SNIR Comparison of OFDM and FBMC. Coherence bandwidth = 108,06 kHz



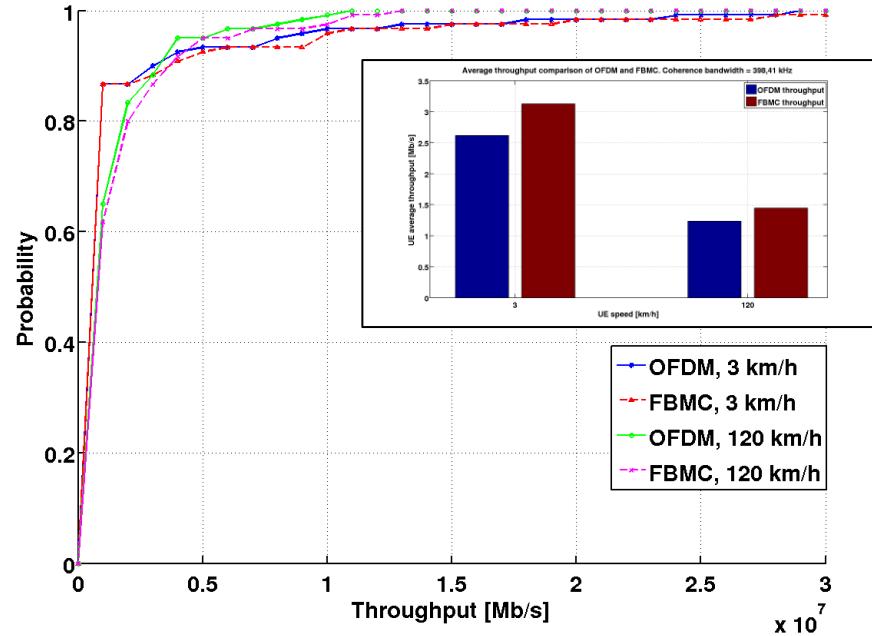
- Difference between the two considered channels : 3GPP EVA (coh.-BW: 398.41 kHz) and (modified) 3GPP ETU (coh.-BW: 108.06 kHz)
- Additional distortion term in FBMC has only influence when coherence bandwidth is smaller than the filter response in frequency domain
- Influence of frequency offsets leads to gain of FBMC at large coherence bandwidth

Simulation Results

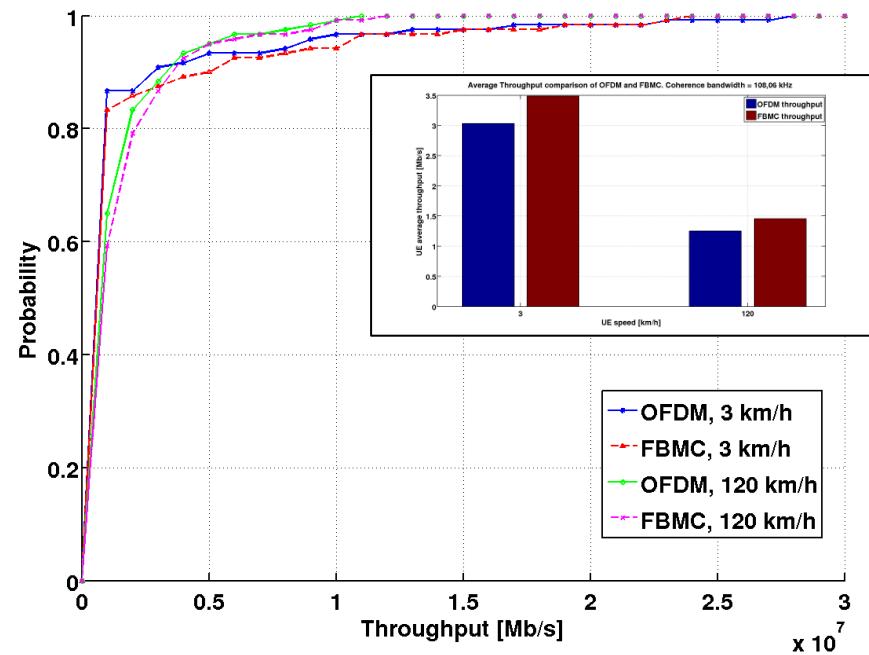
Throughput Comparison

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Throughput-CDF comparison of OFDM and FBMC. Coherence bandwidth = 398,41 kHz



Throughput-CDF comparison of OFDM and FBMC. Coherence bandwidth = 108,06 kHz



- Despite the additional distortion throughput gains through FBMC can be observed in both cases
- Gain through FBMC becomes smaller in ‘frequency selective’ channel

- First steps towards **system level simulations of non-orthogonal waveforms**
 - Quality measure and corresponding effective SNR mapping
 - FBMC frame structure proposed
 - Methods to incorporate frequency offsets
- Incorporate frequency offsets
 - Impact not strong enough to significantly change the performance
 - Time offsets most likely have more significant impact
- FBMC turns out to be more efficient with small frequency selectivity
 - Here the gain of the removed cyclic prefix takes the full effect
 - With larger frequency selectivity the gain decreases.

Thank you for your attention!
www.5gnow.eu

Contact

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Appendix

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Parameter	Value FBMC	Value OFDM	Unit	Abreviation
Usefull bandwidth	10^7	10^7	Hz	B
Number of carriers	1024	1024		N_C
Number of active carriers	601	601		N_{ac}
Cyclic Prefix	N/A	72	carriers	
Carrier spacing	15×10^3	15×10^3	Hz	Δ_f
Sampling rate	15.36×10^6	15.36×10^6	Hz	f_s
Sampling period	65.1×10^{-9}	65.1×10^{-9}	s	$T_S = 1/f_s$
Number of preamble symb. in the subframe	4	4		
Number of data symbols in the subframe	16	9		L
Subframe duration	1.63×10^{-3}	0.93×10^{-3}	s	L_{frame}
Overlapping factor	4	N/A		K
Number of Resource Blocks	50	50		