

Timing Driven Power Gating in High-Level Synthesis

Shih-Hsu Huang and Chun-Hua Cheng

Department of Electronic Engineering
Chung Yuan Christian University, Taiwan

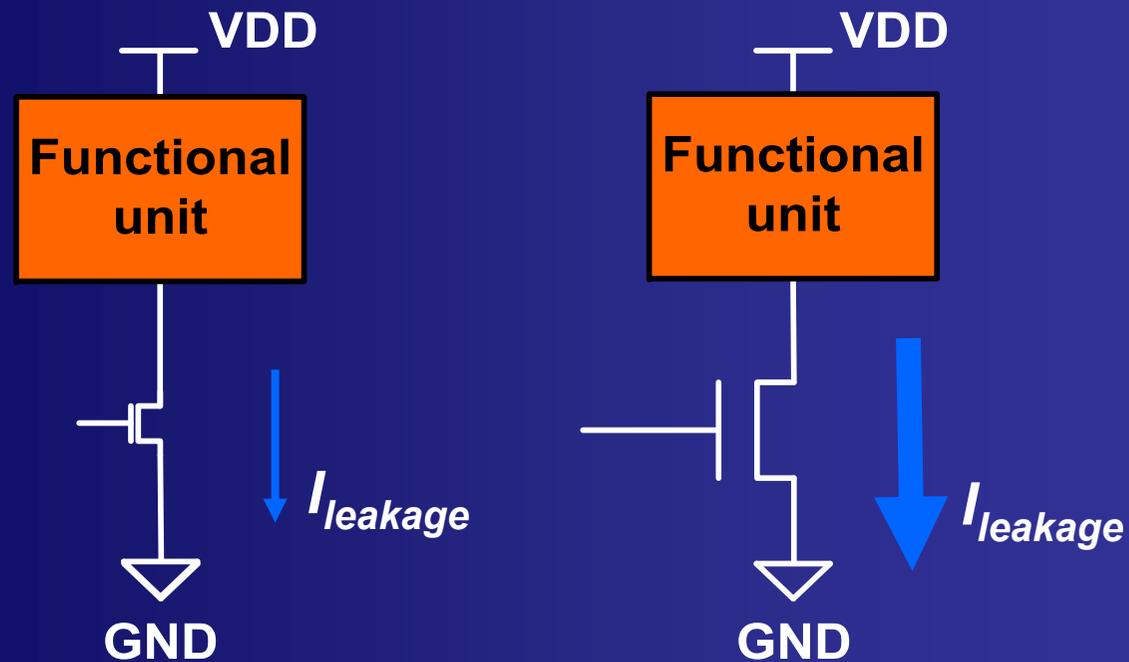


Outline

- Introduction
- Motivation
- Our Approach
 - MILP
 - Heuristic Algorithm
- Experimental Result
- Conclusion

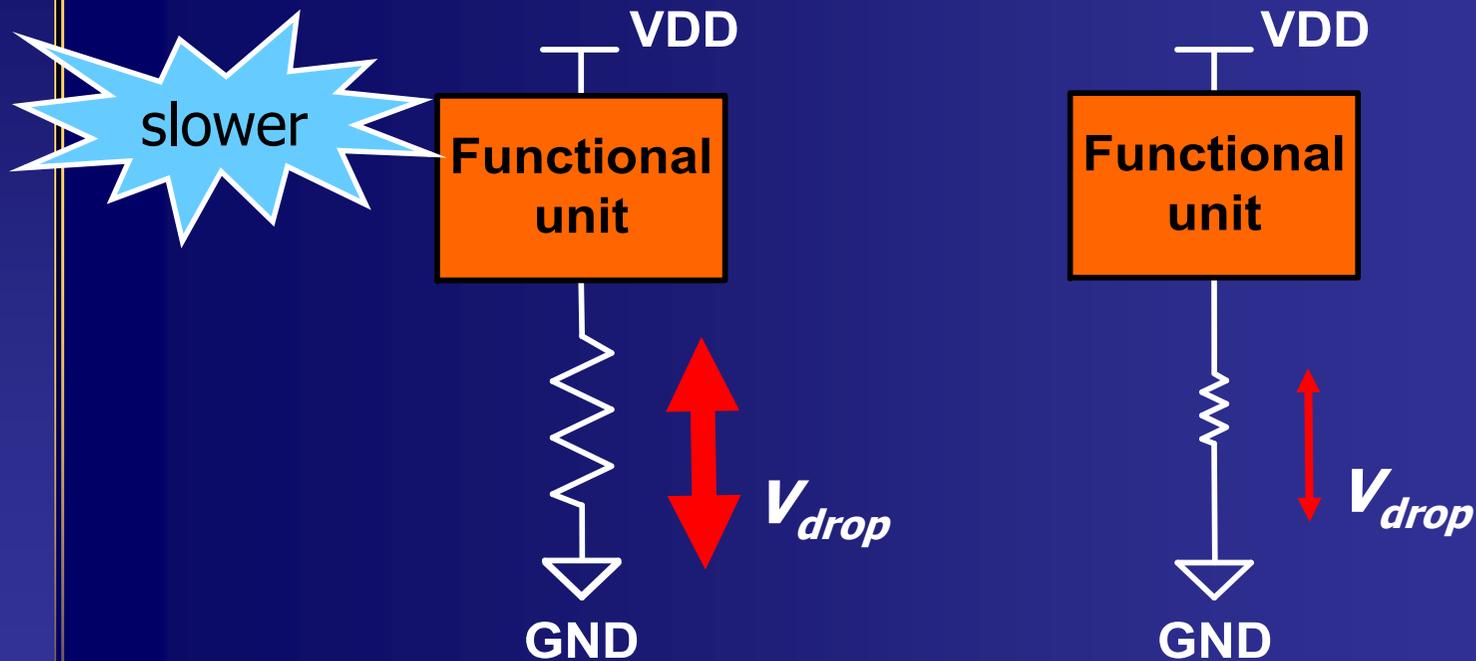
Power Gating Technique (1/2)

- In standby mode : (the sleep transistor is turned off)
 - The standby leakage current of the functional unit is proportional to the size of the sleep transistor.
 - Small sleep transistor to reduce leakage



Power Gating Technique (2/2)

- In active mode : (the sleep transistor is turned on and works as a resistor)
 - The sleep transistor produces a voltage drop that degrades the speed of the functional unit.



Pervious Works

- Up to now, the impact of high-level synthesis on the maximum allowable delays of functional units (for a target clock period) has not been studied.
 - Since the clock skew is assumed to be zero, the maximum allowable delay of each functional unit is definitely the target clock period; thus, there is no need to study this problem.
- However, in modern high-speed circuit design, the clock skew is often intentionally utilized to improve the circuit performance.

Our Contributions

- In this paper, we present the first work to formally draw up the timing driven power gating problem in the high-level synthesis of non-zero clock skew circuits.
 - Given a target clock period and design constraints, our objective is to derive the minimum-standby-leakage-current resource binding solution.
- Our work includes the following two aspects
 - First, we propose an MILP (mixed integer linear programming) approach to guarantee obtaining the optimal solution.
 - Second, we also propose a heuristic approach to deal with the same problem in polynomial time complexity.

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Functional Units with Power Gating

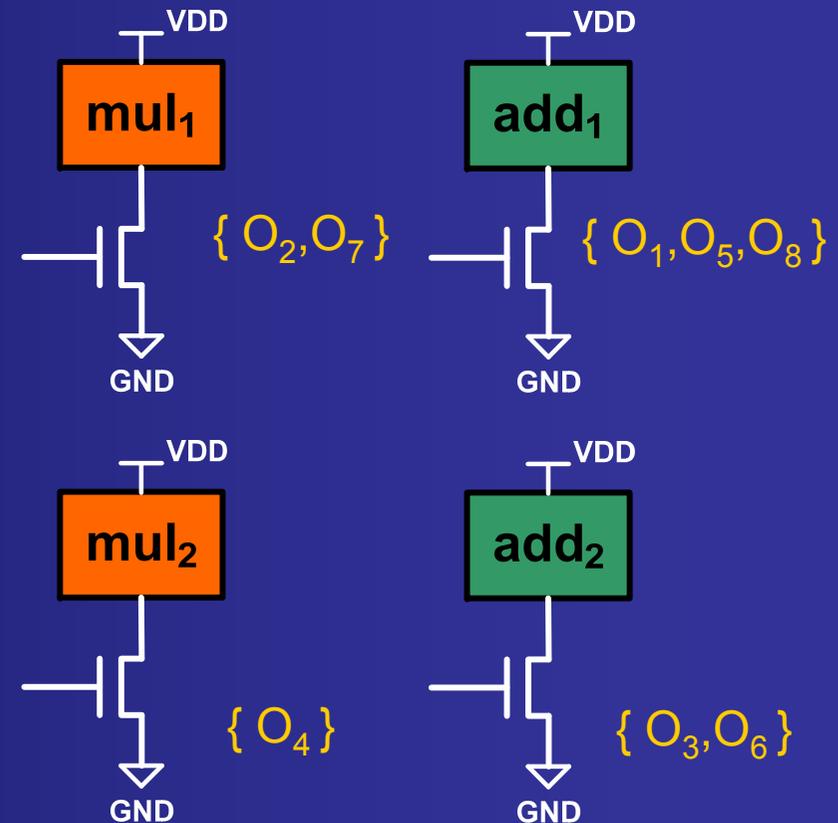
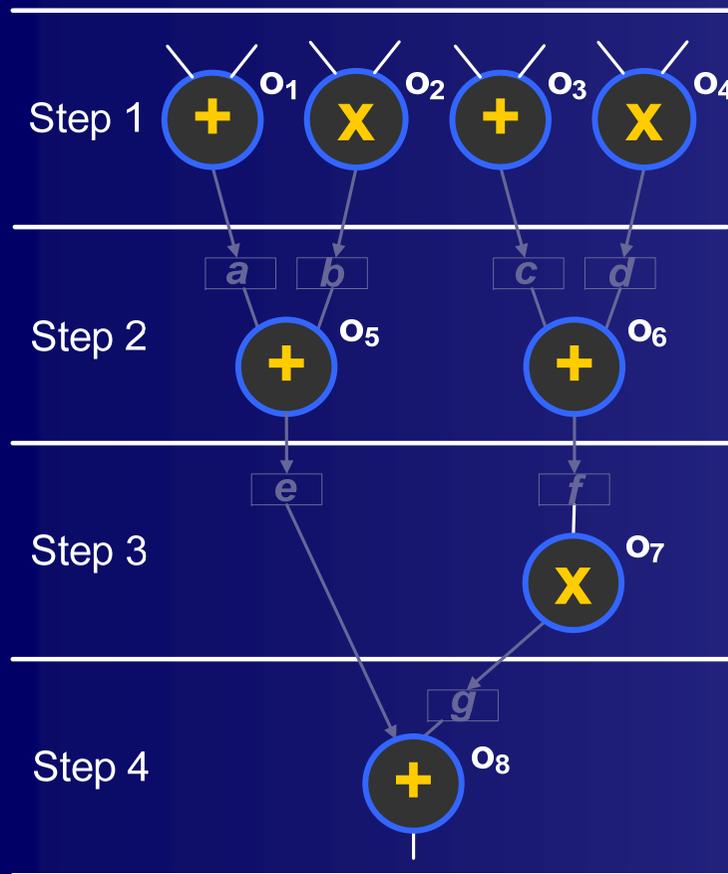
Suppose we are given two multipliers, called mul_1 and mul_2 , and two adders, called add_1 and add_2 :

Type of Functional Unit	Transistor Size	Delay (min,max)	Leakage
Multiplier (mul)	Small (S)	(34,40)	5
	Medium (M)	(33,38)	20
	Large (L)	(28,34)	35
Adder (add)	Small (S)	(10,12)	4
	Large (L)	(8,10)	5

If the power gating implementation selection is mul_1 (fast), mul_2 (fast), add_1 (fast), and add_2 (fast)
Total standby leakage current is 80 (due to $35+35+5+5$)

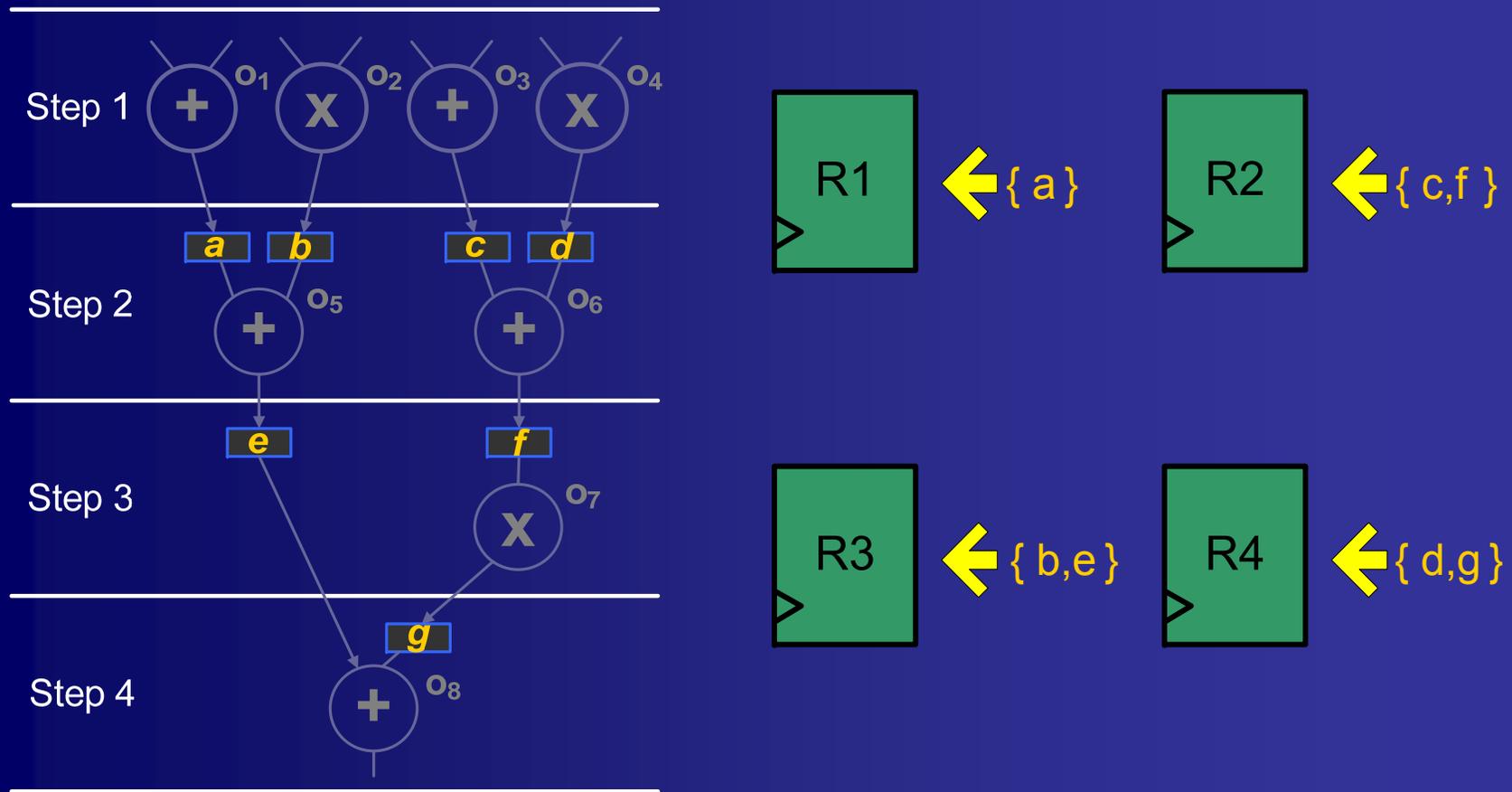
Resource Binding (Functional Unit)

- Suppose we are given two multipliers, called mul_1 and mul_2 , and two adders, called add_1 and add_2 :



Resource Binding (Register)

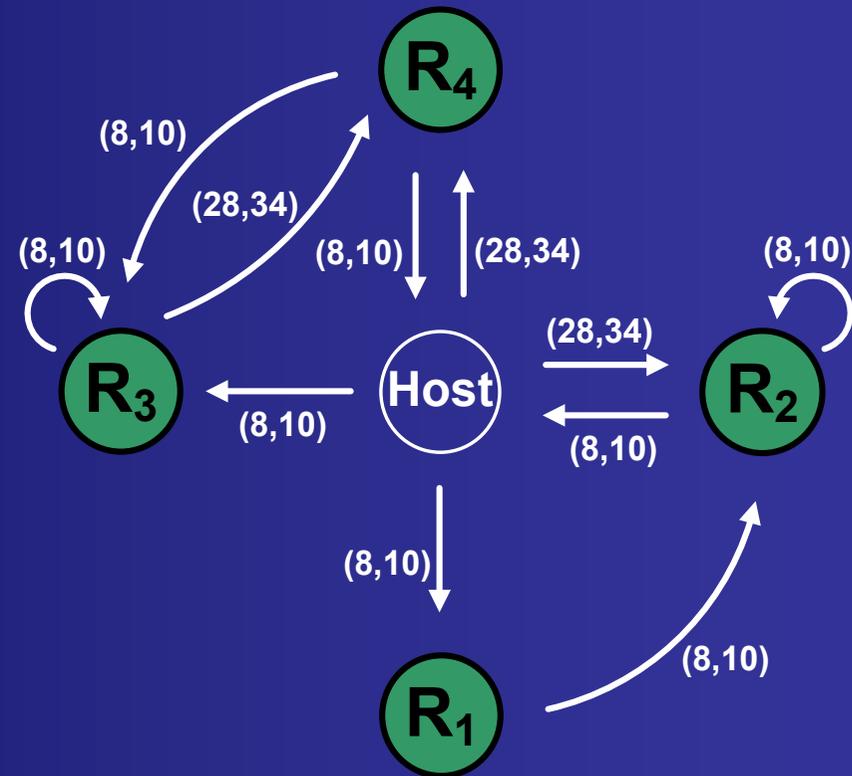
- Suppose we are given four registers, called R1, R2, R3, and R4:



Circuit Graph of Resource Binding

$\text{mul}_1(\text{fast}) = \{o_2, o_7\}$,
 $\text{mul}_2(\text{fast}) = \{o_4\}$,
 $\text{add}_1(\text{fast}) = \{o_1, o_5, o_8\}$,
 $\text{add}_2(\text{fast}) = \{o_3, o_6\}$,
 $R1 = \{a\}$, $R2 = \{b, e\}$,
 $R3 = \{c, f\}$, and $R4 = \{d, g\}$.

Resource Binding



Circuit Graph

If the clock skew is zero,
the clock period cannot be less than 34.

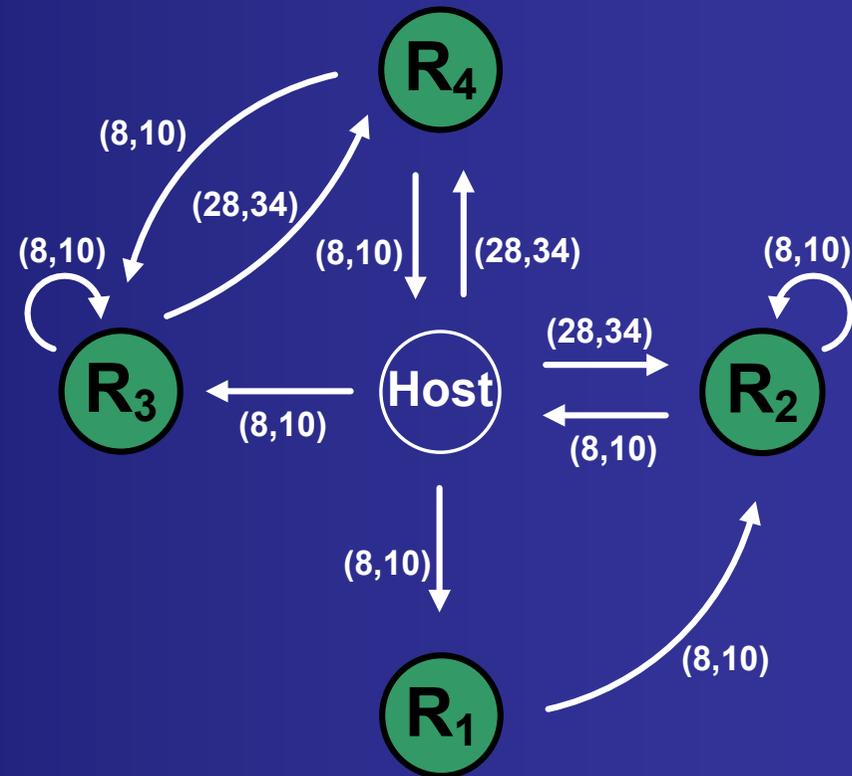
Min-Period Clock Skew Scheduling

- By properly scheduling the clock arrival time of registers, the clock period can be shorter than the longest combinational delay.
- Several graph-based algorithms use the constraint graph to solve the optimal clock skew scheduling.

DRAWBACK OF EXISTING FLOW

$\text{mul}_1(\text{fast}) = \{o_2, o_7\}$,
 $\text{mul}_2(\text{fast}) = \{o_4\}$,
 $\text{add}_1(\text{fast}) = \{o_1, o_5, o_8\}$,
 $\text{add}_2(\text{fast}) = \{o_3, o_6\}$,
 $R1 = \{a\}$, $R2 = \{b, e\}$,
 $R3 = \{c, f\}$, and $R4 = \{d, g\}$.

Resource Binding



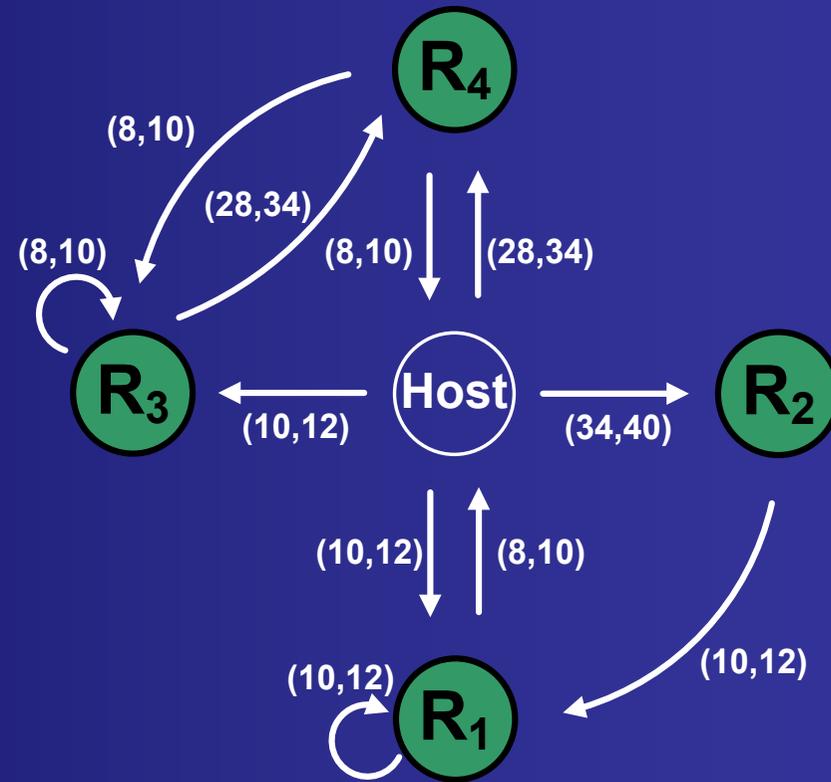
Circuit Graph

The smallest feasible clock period is 22.
Total standby leakage current is 80.

Our Solution

$\text{mul}_1(\text{fast}) = \{o_4, o_7\}$,
 $\text{mul}_2(\text{slow}) = \{o_2\}$,
 $\text{add}_1(\text{fast}) = \{o_3, o_6, o_8\}$,
 $\text{add}_2(\text{slow}) = \{o_1, o_5\}$,
 $R1 = \{a, e\}$, $R2 = \{b\}$,
 $R3 = \{c, f\}$, and $R4 = \{d, g\}$.

Resource Binding



Circuit Graph

The smallest feasible clock period is only 22.
Total standby leakage current is only 49.

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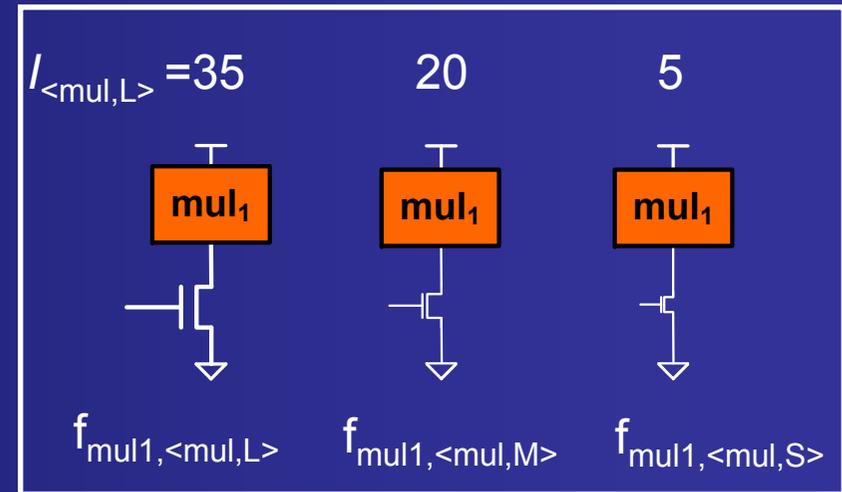
Objective Function

- Our objective function is to minimize total standby leakage current.

- Minimize

$$\sum_{z \in Q} \sum_{w \in h(e(z))} f_{z, \langle e(z), w \rangle} \cdot I_{\langle e(z), w \rangle}$$

Minimize



$$\begin{aligned} & f_{mul1, \langle mul, L \rangle} \times 35 + f_{mul1, \langle mul, M \rangle} \times 20 + f_{mul1, \langle mul, S \rangle} \times 5 + \\ & f_{mul2, \langle mul, L \rangle} \times 35 + f_{mul2, \langle mul, M \rangle} \times 20 + f_{mul2, \langle mul, S \rangle} \times 5 + \\ & f_{add1, \langle add, L \rangle} \times 5 + f_{add1, \langle add, S \rangle} \times 4 + f_{add2, \langle add, L \rangle} \times 5 + \\ & f_{add2, \langle add, S \rangle} \times 4. \end{aligned}$$

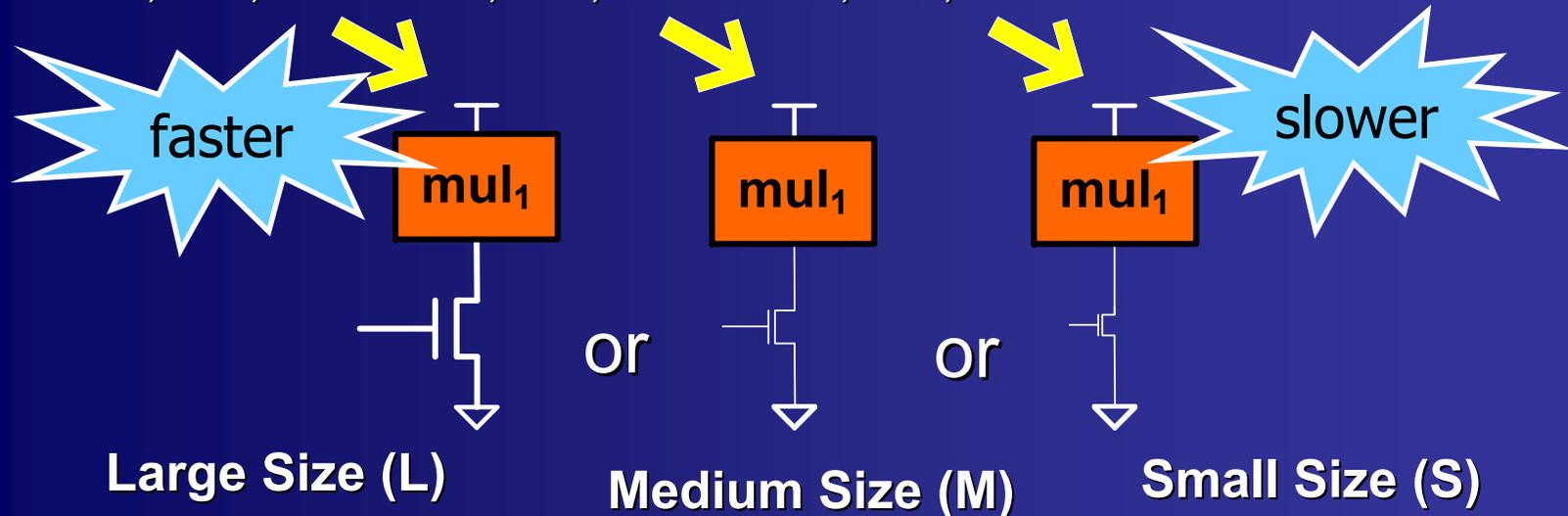
Formula 2

- Each functional unit must select one implementation. Thus, for each functional unit z , we have the following constraint:

$$\sum_{w \in h(e(z))} f_{z, \langle e(z), w \rangle} = 1.$$

- In this example, we have the following constraints:

$$f_{\text{mul}_1, \langle \text{mul}, L \rangle} + f_{\text{mul}_1, \langle \text{mul}, M \rangle} + f_{\text{mul}_1, \langle \text{mul}, S \rangle} = 1; \text{ and so on.}$$



Formula 3

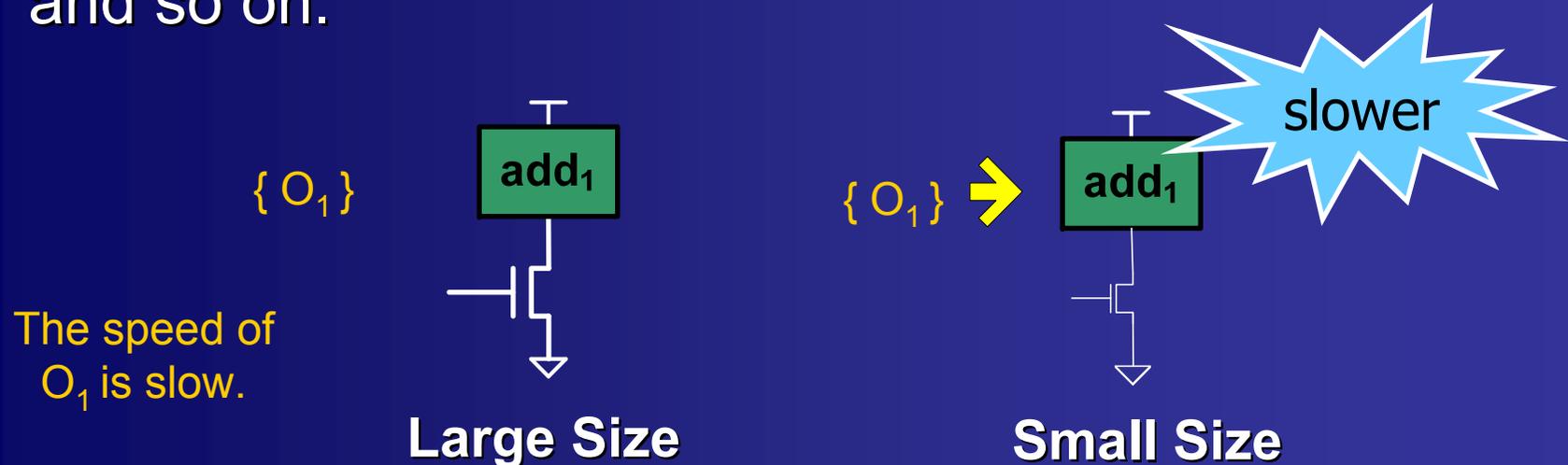
- If the functional unit z is not implemented by $\langle e(z), w \rangle$, then the value of binary variable $y_{j,z,\langle e(z), w \rangle}$ is definitely to be 0. Therefore, we have

the following constraint: $y_{j,z,\langle e(z), w \rangle} \leq f_{z,\langle e(z), w \rangle}$.

- In this example, we have the following constraints:

$$y_{o1,add1,\langle add,L \rangle} \leq f_{add1,\langle add,L \rangle};$$

and so on.



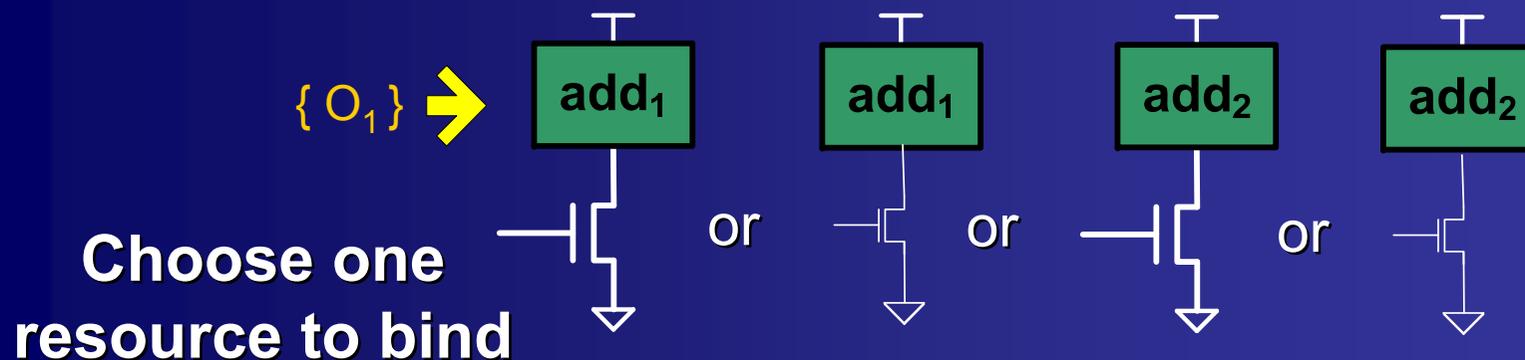
Formula 4

- Each operation must be assigned to one functional unit. Therefore, for each operation o_j , we have the following constraint:

$$\sum_{z \in c(g(j))} \sum_{w \in h(g(j))} y_{j,z,<e(z),w>} = 1.$$

- In this example, we have the following constraints: $y_{o_1, \text{add}_1, <\text{add}, \text{L}>} + y_{o_1, \text{add}_1, <\text{add}, \text{S}>} +$

$y_{o_1, \text{add}_2, <\text{add}, \text{L}>} + y_{o_1, \text{add}_2, <\text{add}, \text{S}>} = 1$; and so on.



Formula 5

- If two operations have overlapping lifetimes, they cannot share the same functional unit. Thus, we have the following constraint:

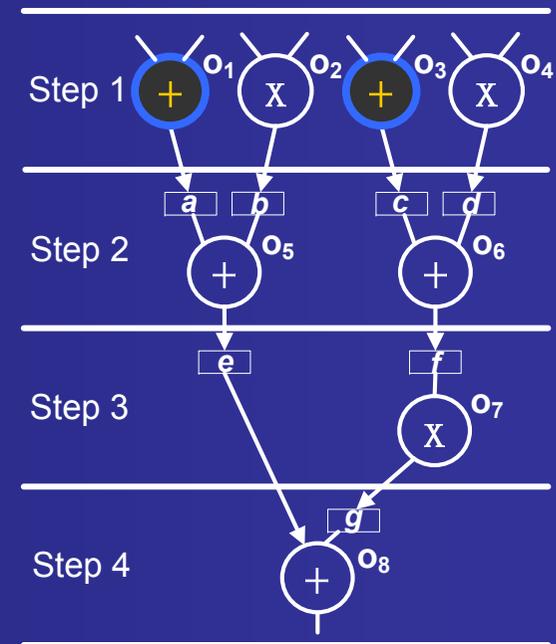
$$y_{j,z,<e(z),w>} + y_{k,z,<e(z),w>} \leq 1.$$

- In this example, we have the following constraints:

$$y_{o1,add1,<add,L>} + y_{o3,add1,<add,L>} \leq 1;$$

$$y_{o1,add1,<add,S>} + y_{o3,add1,<add,S>} \leq 1;$$

and so on.



O₁, O₃ Life Time conflict

Formula 6

- Let P be a constant that denotes the target clock period. For the input variable u and the output variable v of operation o_j , the maximum allowable delay must satisfy the setup constraint:

$$\sum_{z \in c(g(j))} \sum_{w \in h(g(j))} y_{j,z,\langle e(z),w \rangle} \cdot D_{\langle e(z),w \rangle} \leq P - T_u + T_v.$$

- In this example, we have the following constraints:

$$y_{o1,add1,\langle add,L \rangle} \times 10 + y_{o1,add1,\langle add,S \rangle} \times 12 + \\ y_{o1,add2,\langle add,L \rangle} \times 10 + y_{o1,add2,\langle add,S \rangle} \times 12 \\ \leq P - T_{host} + T_a; (P = 22)$$

and so on.



Formula 7

- For the input variable u and the output variable v of operation o_j , the minimum allowable delay must satisfy the hold constraint:

$$T_v - T_u \leq \sum_{z \in c(g(j))} \sum_{w \in h(g(j))} y_{j,z,\langle e(z),w \rangle} \cdot d_{\langle e(z),w \rangle}$$

- In this example, we have the following constraints:

$$T_a - T_{\text{host}} \leq y_{o1,\text{add1},\langle \text{add,L} \rangle} \times 8 + y_{o1,\text{add1},\langle \text{add,S} \rangle} \times 10 +$$

$$y_{o1,\text{add2},\langle \text{add,L} \rangle} \times 8 + y_{o1,\text{add2},\langle \text{add,S} \rangle} \times 10;$$

and so on.



- Formula 8~11 (Use the Register Binding Approach in [7])

Outline

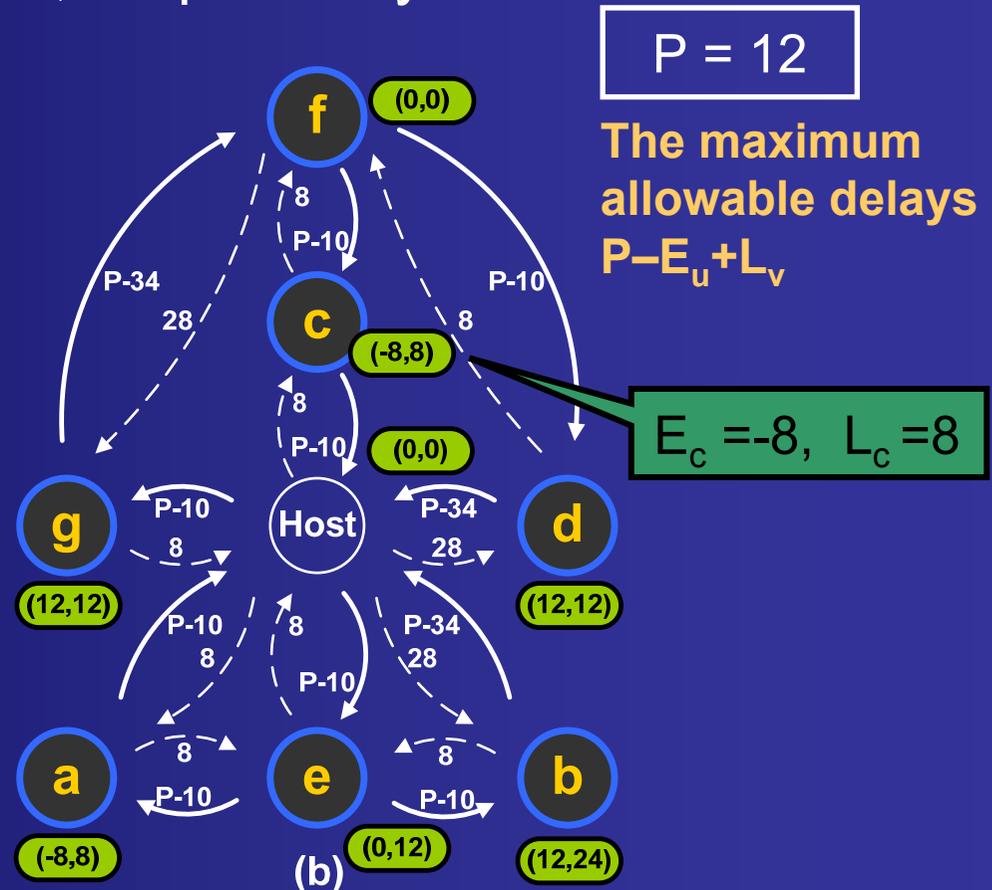
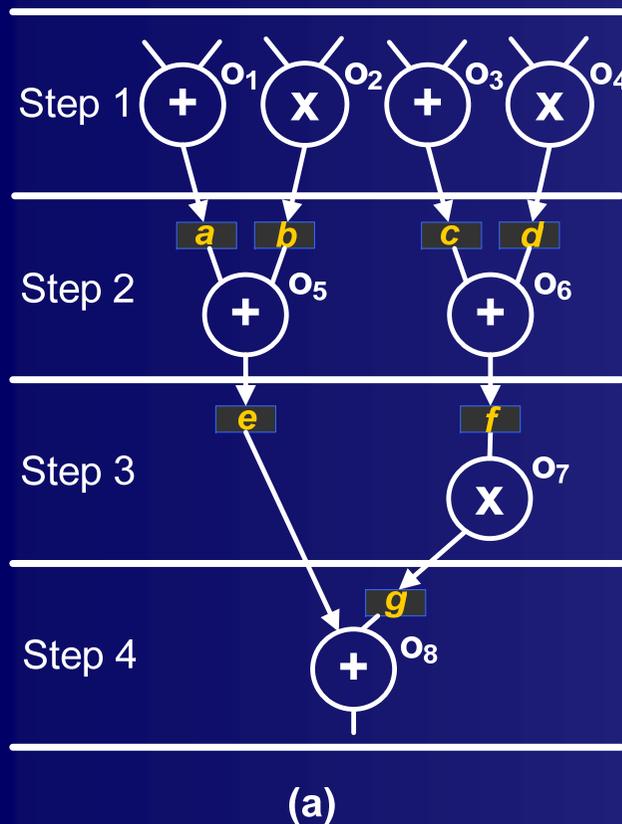
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Heuristic Approach

- We provide a method to estimate the maximum allowable delay of each operation before the resource binding.
- We try to assign the operations that have similar maximum allowable delay to the same functional unit.
- As a result, we can have more non-critical functional units.

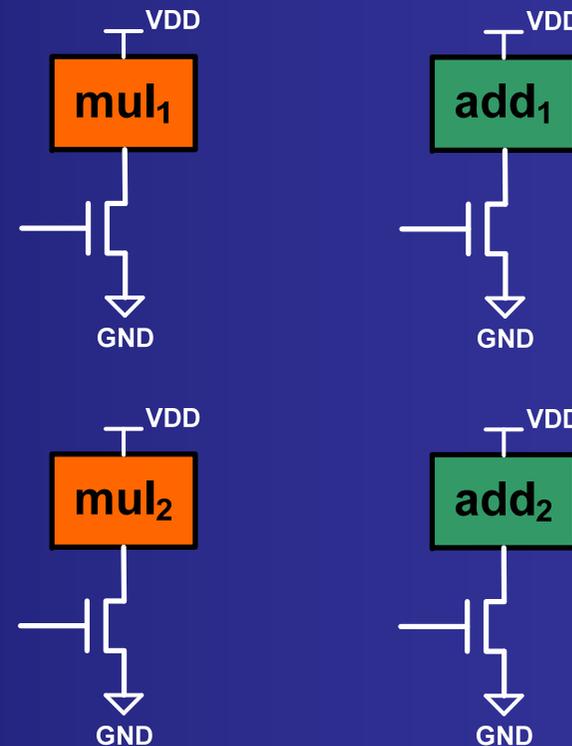
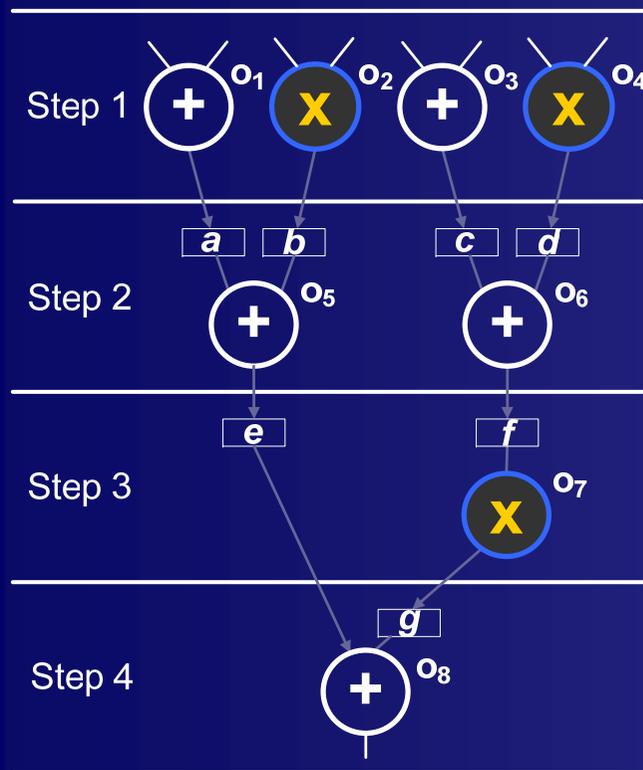
Step 1: Estimate the Maximum Allowable Delay

- The estimated maximum allowable delays of operations $o_1, o_2, o_3, o_4, o_5, o_6, o_7,$ and o_8 are 30, 46, 30, 34, 22, 10, 34, and 10, respectively.



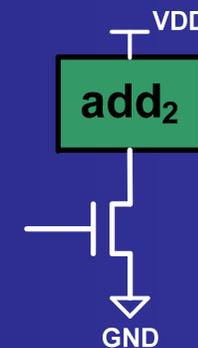
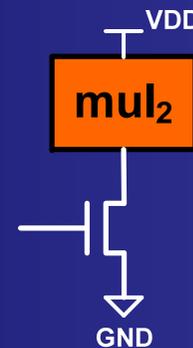
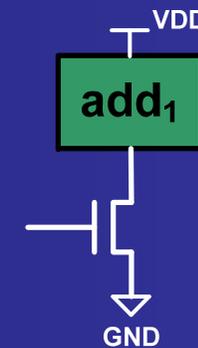
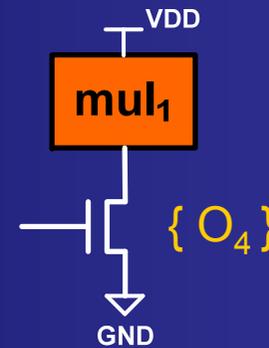
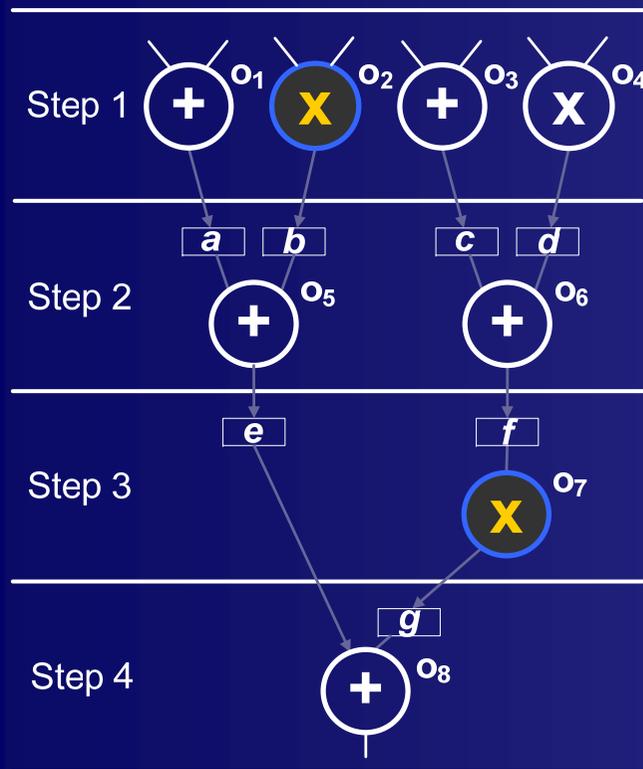
Step 2: Power Gating Implementation (1/5)

- We consider the multiplier type. The estimated maximum allowable delays of operations o_2 , o_4 , and o_7 are 46, 34, and 34, respectively. (o_4 , o_7 high priority)



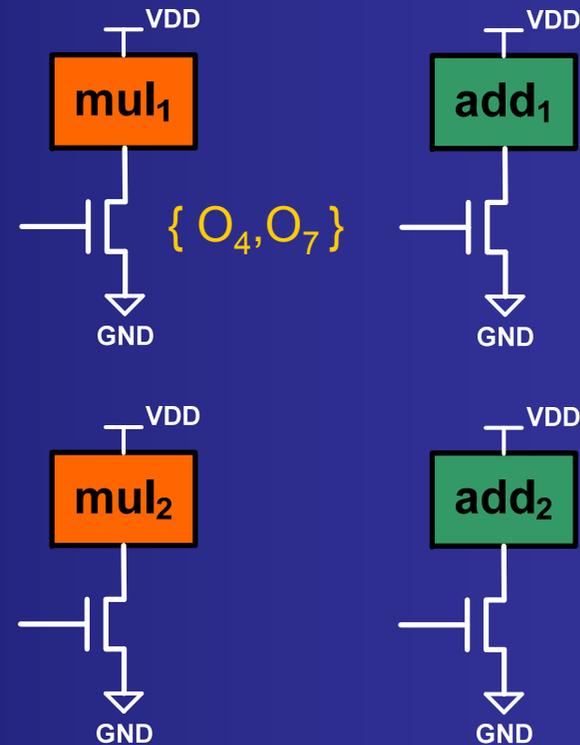
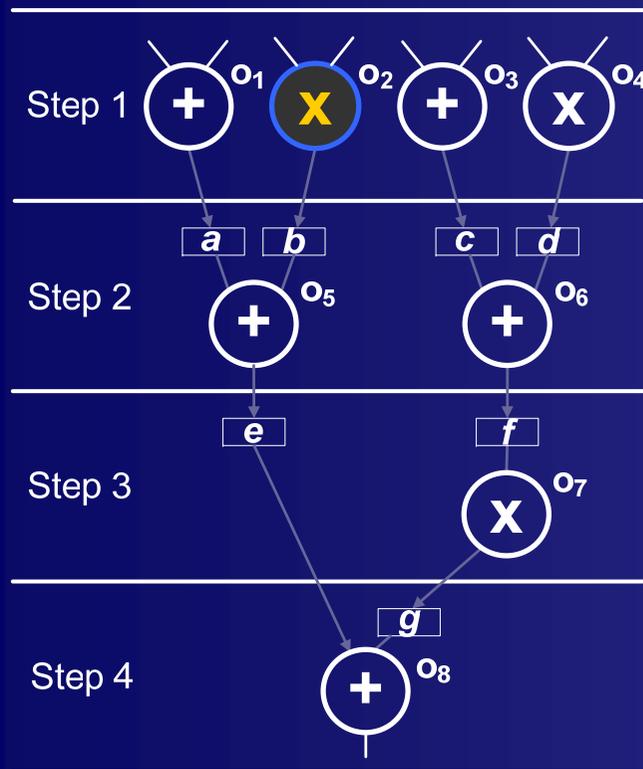
Step 2: Power Gating Implementation (2/5)

- Suppose operation o_4 is chosen. We assign operation o_4 to mul_1 (fast). The estimated maximum allowable delays of operation o_7 is 34.



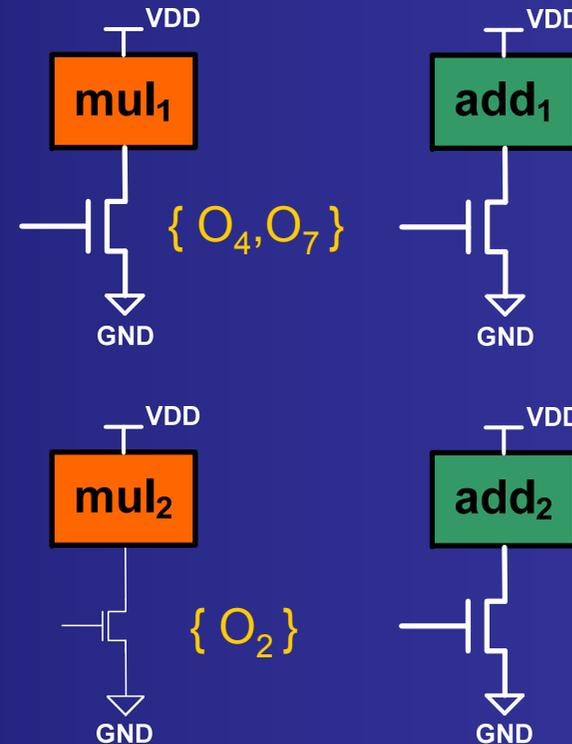
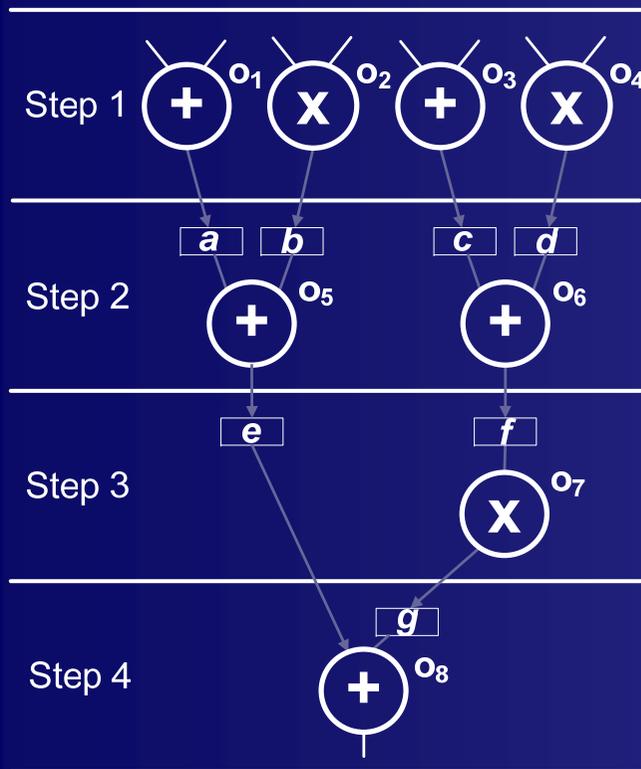
Step 2: Power Gating Implementation (3/5)

- We assign operation o_7 to mul_1 (fast). The estimated maximum allowable delays of operation o_7 is 34.



Step 2: Power Gating Implementation (4/5)

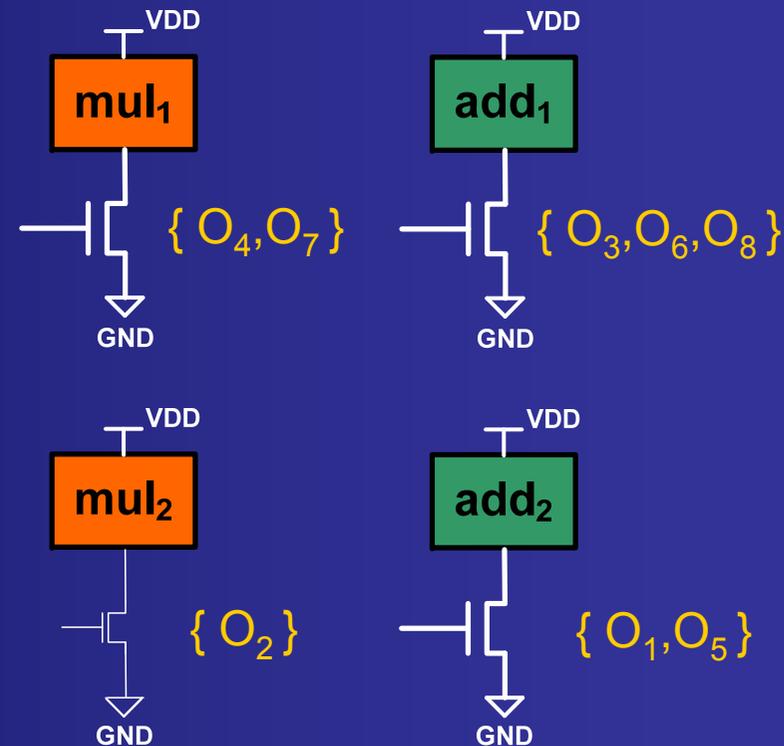
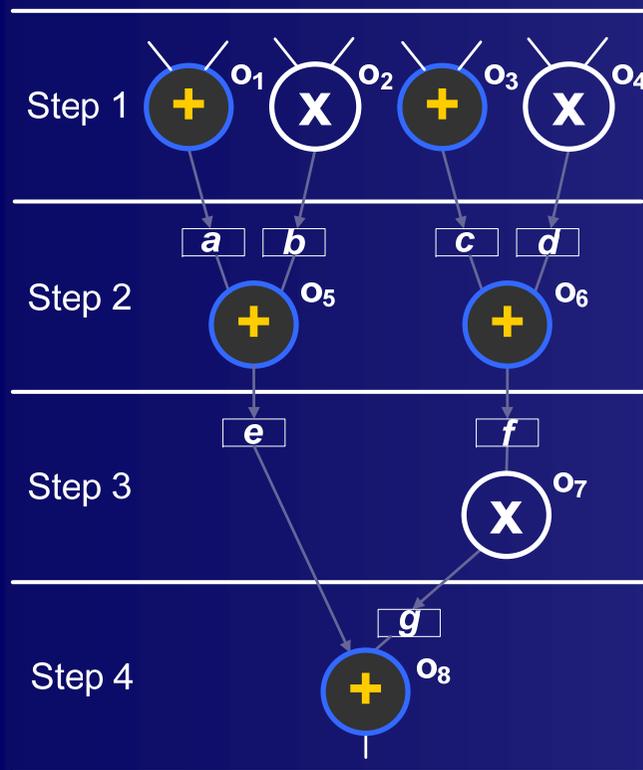
- Because of the lifetime constraint, we find operation o_2 cannot be assigned to the functional unit mul_1 . We assign operation o_2 to mul_2 (slow).



o_2 maximum allowable delays = 46.

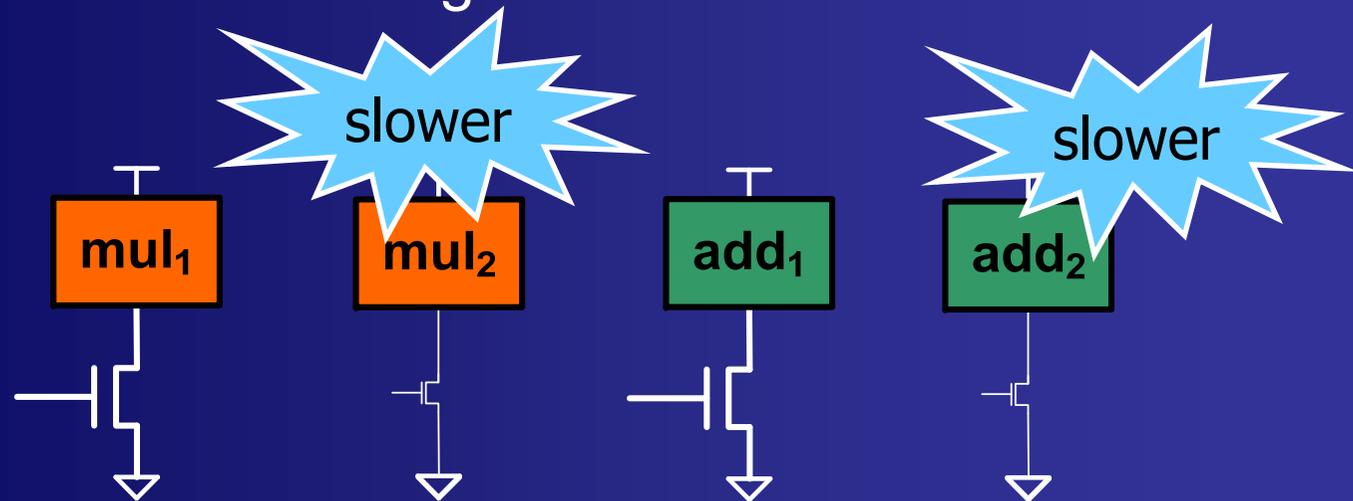
Step 2: Power Gating Implementation (5/5)

- Similarly, we consider the adder type. We have $\text{add}_1(\text{fast}) = \{o_3, o_6, o_8\}$, $\text{add}_2(\text{slow}) = \{o_1, o_5\}$.



Step 4: Adjust the Power Gating Implementation

- Now, the actual maximum allowable delay of each functional unit can be calculated based on the target clock period and the clock skew schedule derived in the third step.
- We adjust the power gating implementation of each functional unit according to its actual maximum allowable delay.



The total standby leakage current is only 49.

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Experimental Environment

- Our platform is Windows XP operating system running on AMD K8-4200+ processor.
 - We use Extended-LINGO Release 10.0 as the mixed integer linear program solver and use C programming language to implement the process of solution space reduction.
- We compare our approach with existing design flow.

Circuit	Improvement				
	Existing Flow	Our MILP	Our Heuristic	Our MILP	Our Heuristic
HAL	1.764	0.484	0.484	72.56%	72.56%
AR	2.644	0.940	1.152	64.45%	56.43%
BF	1.336	0.484	0.484	63.77%	63.77%
EWF	1.350	0.498	0.738	63.11%	45.33%
IDCT1	3.585	1.458	1.956	59.33%	45.44%
Motion	7.376	2.607	4.118	64.65%	54.62%

64.65%

54.62%

Concluding Remarks

- This paper presents the first work to formally formulate the timing driven power gating in the high-level synthesis of a non-zero clock skew circuit.
- Given a target clock period and design constraints, our goal is to find a resource binding solution so that the total standby leakage current is minimized.
- Compared with the existing design flow, our approach has a significant improvement without any overhead.

Thank you