Massive MIMO for Next Generation Wireless Systems

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ABSTRACT

Multi-user MIMO offers big advantages over conventional point-to-point MIMO: it works with cheap single-antenna terminals, a rich scattering environment is not required, and resource allocation is simplified because every active terminal utilizes all of the time-frequency bins. However, multi-user MIMO, as originally envisioned, with roughly equal numbers of service antennas and terminals and frequency-division duplex operation, is not a scalable technology. Massive MIMO (also known as large-scale antenna systems, very large MIMO, hyper MIMO, full-dimension MIMO, and ARGOS) makes a clean break with current practice through the use of a large excess of service antennas over active terminals and time-division duplex operation. Extra antennas help by focusing energy into ever smaller regions of space to bring huge improvements in throughput and radiated energy efficiency. Other benefits of massive MIMO include extensive use of inexpensive low-power components, reduced latency, simplification of the MAC layer, and robustness against intentional jamming. The anticipated throughput depends on the propagation environment providing asymptotically orthogonal channels to the terminals, but so far experiments have not disclosed any limitations in this regard. While massive MIMO renders many traditional research problems irrelevant, it uncovers entirely new problems that urgently need attention: the challenge of making many low-cost low-precision components that work effectively together, acquisition and synchronization for newly joined terminals, the exploitation of extra degrees of freedom provided by the excess of service antennas, reducing internal power consumption to achieve total energy efficiency reductions, and finding new deployment scenarios. This article presents an overview of the massive MIMO concept and contemporary research on the topic.

GOING LARGE: MASSIVE MIMO

Massive multiple-input multiple-output (MIMO) is an emerging technology that scales up MIMO by possibly orders of magnitude compared to the current state of the art. In this article, we follow up on our earlier exposition [1], with a focus on the developments in the last three years; most particularly, energy efficiency, exploitation of excess degrees of freedom, time-division duplex (TDD) calibration, techniques to combat pilot contamination, and entirely new channel measurements.

With massive MIMO, we think of systems that use antenna arrays with a few hundred antennas simultaneously serving many tens of terminals in the same time-frequency resource. The basic premise behind massive MIMO is to reap all the benefits of conventional MIMO, but on a much greater scale. Overall, massive MIMO is an enabler for the development of future broadband (fixed and mobile) networks, which will be energy-efficient, secure, and robust, and will use the spectrum efficiently. As such, it is an enabler for the future digital society infrastructure that will connect the Internet of people and Internet of Things with clouds and other network infrastructure. Many different configurations and deployment scenarios for the actual antenna arrays used by a massive MIMO system can be envisioned (Fig. 1). Each antenna unit would be small and active, preferably fed via an optical or electric digital bus.

Massive MIMO relies on spatial multiplexing, which in turn relies on the base station having good enough channel knowledge, on both the uplink and the downlink. On the uplink, this is easy to accomplish by having the terminals send pilots, based on which the base station estimates the channel responses, quantizes the thus obtained estimates, and feed them back to the base station. The downlink is more difficult. In conventional MIMO systems such as the Long Term Evolution (LTE) standard, the base station sends out pilot waveforms, based on which the terminals estimate the channel responses, quantize the thus obtained estimates, and feed them back to the base station. This will not be feasible in massive MIMO systems, at least not when operating in a high-mobility environment, for two reasons. First, optimal downlink pilots should be mutually orthogonal between the antennas. This means that the amount of time-frequency resources needed for downlink pilots scales with the number of antennas, so a massive MIMO system would require up to 100 times more such resources than a conventional system. Second,
the number of channel responses each terminal must estimate is also proportional to the number of base station antennas. Hence, the uplink resources needed to inform the base station of the channel responses would be up to 100 times larger than in conventional systems. Generally, the solution is to operate in TDD mode, and rely on reciprocity between the uplink and downlink channels, although frequency-division duplex (FDD) operation may be possible in certain cases [2].

While the concepts of massive MIMO have been mostly theoretical so far, stimulating much research particularly in random matrix theory and related mathematics, basic testbeds are becoming available [3], and initial channel measurements have been performed [4, 5].

**THE POTENTIAL OF MASSIVE MIMO**

Massive MIMO technology relies on phase-coherent but computationally very simple processing of signals from all the antennas at the base station. Some specific benefits of a massive MU-MIMO system are:

- **Massive MIMO can increase the capacity 10 times or more and simultaneously improve the radiated energy efficiency** on the order of 100 times. The capacity increase results from the aggressive spatial multiplexing used in massive MIMO. The fundamental principle that makes the dramatic increase in energy efficiency possible is that with a large number of antennas, energy can be focused with extreme sharpness into small regions in space (Fig. 2). The underlying physics is coherent superposition of wavefronts. By appropriately shaping the signals sent out by the antennas, the base station can make sure that all wavefronts collectively emitted by all antennas add up constructively at the locations of the intended terminals, but destructively (randomly) almost everywhere else. Interference between terminals can be suppressed even further by using, for example, zero-forcing (ZF). This, however, may come at the cost of more transmitted power, as illustrated in Fig. 2.

More quantitatively, Fig. 3 (from [6]) depicts the fundamental trade-off between energy efficiency in terms of the total number of bits (sum rate) transmitted per Joule per terminal receiving service of energy spent, and spectral efficiency in terms of total number of bits (sum rate) transmitted per unit of radio spectrum consumed. The figure illustrates the relation for the uplink, from the terminals to the base station (the downlink performance is similar). The figure shows the trade-off for three cases:

- **A reference system with one single antenna serving a single terminal (purple)**
- **A system with 100 antennas serving a single terminal using conventional beamforming (green)**
- **A massive MIMO system with 100 antennas simultaneously serving multiple (about 40 here) terminals (red, using maximum ratio combining, and blue, using ZF).**

The attractiveness of maximum ratio combining (MRC) compared with ZF is not only its computational simplicity — multiplication of the received signals by the conjugate channel responses — but also that it can be performed in a distributed fashion, independently at each antenna unit. While ZF also works fairly well for a conventional or moderately sized MIMO system, MRC generally does not. The reason that MRC works so well for massive MIMO is that the channel responses associated with different terminals tend to be nearly orthogonal when the number of base station antennas is large.

The prediction in Fig. 3 is based on an information-theoretic analysis that takes into account intracell interference, as well as the bandwidth and energy cost of using pilots to acquire channel state information in a high-mobility environment [6]. With the MRC receiver, we operate in the nearly noise-limited regime of information theory. This means providing each terminal with a rate of about 1 b/complex dimension (1 b/s/Hz). In a massive MIMO system, when using MRC and operating in the “green” regime (i.e., scaling down the power as much as possible without seriously affecting the overall spectral efficiency), multiuser interference and effects from hardware imperfections tend to be overwhelmed by the thermal noise. The reason that the overall spectral efficiency still can be 10 times higher than in conventional MIMO is that many tens of terminals are served simultaneously, in the same time-frequency resource. When operating in the 1 b/dimension/terminal regime, there is also some evidence that intersymbol interference can be treated as additional thermal noise [7], hence offering a way of disposing with orthogonal frequency-division multiplexing (OFDM) as a means of combating intersymbol interference.

To understand the scale of the capacity gains massive MIMO offers, consider an array consisting of 6400 omnidirectional antennas (total form factor $6400 \times (\lambda/2)^2 \approx 40 \text{ m}^2$) transmitting with a total power of 120 W (i.e., each antenna radiat-
ing about 20 mW) over a 20 MHz bandwidth in the personal communications services (PCS) band (1900 MHz). The array serves 1000 fixed terminals randomly distributed in a disk of radius 6 km centered on the array, each terminal having an 8 dB gain antenna. The height of the antenna array is 30 m, and the height of the terminals is 5 m. Using the Hata-COST231 model, we find that the path loss is 127 dB at 1 km range, and the range-decay exponent is 3.52. There is also log-normal shadow fading with 8 dB standard deviation. The receivers have a 9 dB noise figure. One-quarter of the time is spent on transmission of uplink pilots for TDD channel estimation, and it is assumed that the channel is substantially constant over intervals of 164 ms in order to estimate the channel gains with sufficient accuracy. Downlink data is transmitted via maximum ratio transmission (MRT) beamforming combined with power control, where the 5 percent of terminals with the worst channels are excluded from service. We use a capacity lower bound from [8] extended to accommodate slow fading, near/far effects and power control, which accounts for receiver noise, channel estimation errors, the overhead of pilot transmission, and the imperfections of MRT beamforming. We use optimal max-min power control, which confers an equal signal-to-interference-plus-noise ratio on each of the 950 terminals and therefore equal throughput. Numerical averaging over random terminal locations and the shadow fading shows that 95 percent of the terminals will receive a throughput of 21.2 Mb/s/terminal. Overall, the array in this example will offer the 1000 terminals a total downlink throughput of 20 Gb/s, resulting in a sum-spectral efficiency of 1000 b/s/Hz. This would be enough, for example, to provide 20 Mb/s broadband service to each of 1000 homes. The max-min power control provides equal service simultaneously to 950 terminals. Other types of power control combined with time-division multiplexing could accommodate heterogeneous traffic demands of a larger set of terminals.

The MRC receiver (for the uplink) and its counterpart MRT precoding (for the downlink) are also known as matched filtering (MF) in the literature.

Massive MIMO is a game-changing technology with regard to theory, systems, and implementation. With massive MIMO, expensive ultra-linear 50 W amplifiers used in conventional systems are replaced by hundreds of low-cost

Figure 2. Relative field strength around a target terminal in a scattering environment of size 800 λ × 800 λ when the base station is placed 1600 λ to the left. Average field strengths are calculated over 10,000 random placements of 400 scatterers when two different linear precoders are used: a) MRT precoders; b) ZF precoders. Left: pseudo-color plots of average field strengths, with target user positions at the center (•) and four other users nearby (○). Right: average field strengths as surface plots, allowing an alternate view of the spatial focusing.
amplifiers with output power in the milli-Watt range. The contrast to classical array designs, which use few antennas fed from high-power amplifiers, is significant. Several expensive and bulky items, such as large coaxial cables, can be eliminated altogether. (The typical coaxial cables used for tower-mounted base stations today are more than 4 cm in diameter!)

Massive MIMO reduces the constraints on accuracy and linearity of each individual amplifier and RF chain. All that matters is their combined action. In a way, massive MIMO relies on the law of large numbers to make sure that noise, fading, and hardware imperfections average out when signals from a large number of antennas are combined in the air. The same property that makes massive MIMO resilient against fading also makes the technology extremely robust to failure of one or a few of the antenna unit(s).

A massive MIMO system has a large surplus of degrees of freedom. For example, with 200 antennas serving 20 terminals, 180 degrees of freedom are unused. These degrees of freedom can be used for hardware-friendly signal shaping. In particular, each antenna can transmit signals with small peak-to-average ratio [9] or even constant envelope [10] at a very modest penalty in terms of increased total radiated power. Such (near-constant) envelope signaling facilitates the use of extremely cheap and power-efficient RF amplifiers. The techniques in [9, 10] must not be confused with conventional beamforming techniques or equal-magnitude-weight beamforming techniques. This distinction is explained in Fig. 4. With (near) constant-envelope multiuser precoding, no beams are formed, and the signals emitted by each antenna are not formed by weighing a symbol. Rather, a wavefield is created such that when this wavefield is sampled at the spots where the terminals are located, the terminals see precisely the signals we want them to see. The fundamental property of the massive MIMO channel that makes this possible is that the channel has a large nullspace: almost anything can be put into this nullspace without affecting what the terminals see. In particular, components can be put into this nullspace that make the transmitted waves resulting from these multiple paths interfere destructively. It is this fading that makes it hard to build low-latency wireless links. If the terminal is trapped in a fading dip, it has to wait until the propagation channel has sufficiently changed until any data can be received. Massive MIMO relies on the law of large numbers and beamforming in order to avoid fading dips, so fading no longer limits latency.

Massive MIMO enables a significant reduction of latency on the air interface.

The performance of wireless communications systems is normally limited by fading. Fading can render the received signal strength very small at certain times. This happens when the signal sent from a base station travels through multiple paths before it reaches the terminal, and the waves resulting from these multiple paths interfere destructively. It is this fading that makes it hard to build low-latency wireless links. If the terminal is trapped in a fading dip, it has to wait until the propagation channel has sufficiently changed until any data can be received. Massive MIMO relies on the law of large numbers and beamforming in order to avoid fading dips, so fading no longer limits latency.

Massive MIMO simplifies the multiple access layer.

Due to the law of large numbers, the channel hardens so that frequency domain scheduling no longer pays off. With OFDM, each subcarrier in a massive MIMO system will have substantially the same channel gain. Each terminal can be given the whole bandwidth, which renders most of the physical layer control signaling redundant.

Massive MIMO increases the robustness against both unintended man-made interference and intentional jamming.

Intentional jamming of civilian wireless systems is a growing concern and a serious cybersecurity threat that seems to be little known to the public. Simple jammers can be bought off the Internet for a few hundred dollars, and equipment that used to be military-grade can be put together using off-the-shelf software radio-based platforms for a few thousand dollars.
Numerous recent incidents, especially in public safety applications, illustrate the magnitude of the problem. During the EU summit in Gothenburg, Sweden, in 2001, demonstrators used a jammer located in a nearby apartment, and during critical phases of riots, the chief commander could not reach any of the 700 police officers engaged [11].

Due to the scarcity of bandwidth, spreading information over frequency just is not feasible, so the only way of improving robustness of wireless communications is to use multiple antennas. Massive MIMO offers many excess degrees of freedom that can be used to cancel signals from intentional jammers. If massive MIMO is implemented using uplink pilots for channel estimation, smart jammers could cause harmful interference with modest transmission power. However, more clever implementations using joint channel estimation and decoding should be able to substantially diminish that problem.

LIMITING FACTORS OF MASSIVE MIMO

CHANNEL RECIPROCITY

Time-division duplexing operation relies on channel reciprocity. There appears to be a reasonable consensus that the propagation channel itself is essentially reciprocal unless the propagation is affected by materials with strong magnetic properties. However, the hardware chains in the base station and terminal transceivers may not be reciprocal between the uplink and the downlink. Calibration of the hardware chains does not seem to constitute a serious problem, and there are calibration-based solutions that have already been tested to some extent in practice [3, 12]. Specifically, [3] treats reciprocity calibration for a 64-antenna system in some detail and claims a successful experimental implementation.

Note that calibration of the terminal uplink and downlink chains is not required in order to obtain the full beamforming gains of massive MIMO: if the base station equipment is properly calibrated, the array will indeed transmit a coherent beam to the terminal. (There will still be some mismatch within the receiver chain of the terminal, but this can be handled by transmitting pilots through the beam to the terminal; the overhead for these supplementary pilots is very small.) Absolute calibration within the array is not required. Instead, as proposed in [3], one of the antennas can be treated as a reference, and signals can be traded between the reference antenna and each of the other antennas to derive a compensation factor for that antenna. It may be possible to entirely forgo reciprocity calibration within the array: for example if the maximum phase difference between the uplink and downlink chains were less than 60°, coherent beamforming would still occur (at least with MRT beamforming), albeit with a possible 3 dB reduction in gain.

PILOT CONTAMINATION

Ideally, every terminal in a massive MIMO system is assigned an orthogonal uplink pilot sequence. However, the maximum number of orthogonal pilot sequences that can exist is upper-bounded by the duration of the coherence interval divided by the channel delay spread.
Massive MIMO (and especially MRC/MRT processing) relies to a large extent on a property of the radio environment called favorable propagation. Simply stated, favorable propagation means that the propagation channel responses from the base station to different terminals are sufficiently different. To study the behavior of massive MIMO systems, channel measurements have to be performed using realistic antenna arrays. This is so because the channel behavior using large arrays differs from that usually experienced using conventional smaller arrays. The most important differences are that:

- There might be large-scale fading over the array.
- The small-scale signal statistics may also change over the array. Of course, this is also true for physically smaller arrays with directional antenna elements pointing in various directions.

Figure 5 shows pictures of the two massive MIMO arrays used for the measurements reported in this article. On the left is a compact circular massive MIMO array with 128 antenna ports. This array consists of 16 dual-polarized patch antenna elements arranged in a circle, with 4 such circles stacked on top of each other. Besides having the advantage of being compact, this array also provides the possibility to resolve scatters at different elevations, but it suffers from worse resolution in azimuth due to its limited aperture. To the right is a physically large linear (virtual) array, where a single omnidirectional antenna element is moved to 128 different positions in an otherwise static environment to emulate a real array with the same dimensions.

One way of quantifying how different the channel responses to different terminals are is to look at the spread between the smallest and largest singular values of the matrix that contains the channel responses. Figure 6 illustrates this for a case with 4 user terminals and a base station having 4, 32, and 128 antenna ports, respectively, configured as either a physically large single-polarized linear array or a compact dual-polarized circular array. More specifically, the figure shows the cumulative density function (CDF) of the difference between the smallest and largest singular values for the different measured (narrowband) frequency points in the different cases. As a reference, we also show simulated results for ideal independent identically distributed (i.i.d.) channel matrices, often used in theoretical studies. The measurements were performed outdoors in the Lund University campus area. The center frequency was 2.6 GHz and the measurement bandwidth 50 MHz. When using the cylindrical array, the RUSK Lund channel sounder was employed, while a network analyzer was used for the synthetic linear array measurements. The first results from the campaign were presented in [4].

For the 4-element array, the median of the singular value spread is about 23–18 dB. This number is a measure of the fading margin, the additional power that has to be used in order to serve all users with a reasonable received signal.
power. With the massive linear array, the spread is less than 3 dB. In addition, note that none of the curves has any substantial tail. This means that the probability of seeing a singular value spread larger than 3 dB anywhere over the measured bandwidth is essentially negligible.

To further illustrate the influence of different numbers of antenna elements at the base station and antenna configuration, we plot in Fig. 7 the sum rate for 4 closely spaced users (less than 2 m between users at a distance of about 40 m from the base station) in a non line-of-sight (NLOS) scenario when using MRT as precoding. The transmit power is normalized so that on average, the interference-free signal-to-noise ratio at the terminals is 10 dB.

As can be seen in Fig. 7, the sum rate approaches that of the theoretical interference-free case as the number of antennas at the base station increases. The shaded areas in red (for the linear array) and blue (for the circular array) shows the 90 percent confidence intervals of the sum rates for the different narrowband frequency realizations. As before, the variance of the sum rate decreases as the number of antennas increases, but slowly for the measured channels. The slow decrease can, at least partially, be attributed to the shadow fading occurring across the arrays: for the linear array in the form of shadowing by external objects along the array, and for the cylindrical array in the form of shadowing caused by directive antenna elements pointing in the wrong direction. The performance of the physically large array approaches that of the theoretical i.i.d. case when the number of antennas grows large. The compact circular array has inferior performance compared to the linear array due to its smaller aperture — it cannot resolve the scatterers as well as the physically large array — and its directive antenna elements sometimes pointing in the wrong direction. Also, due to the fact that most of the scatterers are seen at the same horizontal angle, the possibility to resolve scatterers at different elevations gives only marginal contributions to the sum rate in this scenario.

It should be mentioned here that when using somewhat more complex, but still linear, precoding methods such as ZF or minimum mean square error, the convergence to the i.i.d. channel performance is faster and the variance of the sum rate lower as the number of base station antennas grows; see [4] for further details. Also, another aspect worth mentioning is that for a very tricky propagation scenario, such as closely spaced users in line-of-sight conditions, it seems that the large array is able to separate the users to a reasonable extent using the different spatial signatures the users have at the base station due to the enhanced spatial resolution. This would not be possible with conventional MIMO. These conclusions are also in line with the observations in [5], where another outdoor measurement campaign is described and analyzed.
Overall, there is compelling evidence that the assumptions on favorable propagation underpinning massive MIMO are substantially valid in practice. Depending on the exact configuration of the large array and the precoding algorithms used, the convergence toward the ideal performance may be faster or slower as the number of antennas is increased. However, with about 10 times more base station antennas than the number of users, it seems that it is possible to get stable performance not far from the theoretically ideal performance also under what are normally considered very difficult propagation conditions.

**MASSIVE MIMO: A GOLD MINE OF RESEARCH PROBLEMS**

While massive MIMO renders many traditional problems in communication theory less relevant, it uncovers entirely new problems that need research.

**Fast and distributed coherent signal processing:** Massive MIMO arrays generate vast amounts of baseband data that must be processed in real time. This processing will have to be simple, and simple means linear or nearly linear. Fundamentally, this is good in many cases (Fig. 3). Much research needs be invested in the design of optimized algorithms and their implementation. On the downlink, there is enormous potential for ingenious precoding schemes. Some examples of recent work in this direction include [19].

**The challenge of low-cost hardware:** Building hundreds of RF chains, up/down converters, analog-to-digital (A/D)-digital-to-analog (D/A) converters, and so forth, will require economy of scale in manufacturing comparable to what we have seen for mobile handsets.

**Hardware impairments:** Massive MIMO relies on the law of large numbers to average out noise, fading and to some extent, interference. In reality, massive MIMO must be built with low-cost components. This is likely to mean that hardware imperfections are larger: in particular, phase noise and I/Q imbalance. Low-cost and power-efficient A/D converters yield higher levels of quantization noise. Power amplifiers with very relaxed linearity requirements will necessitate the use of per-antenna low peak-to-average signaling, which, as already noted, is feasible with a large excess of transmitter antennas. With low-cost phase locked loops or even free-running oscillators at each antenna, phase noise may become a limiting factor. However, what ultimately matters is how much the phase will drift between the point in time when a pilot symbol is received and the point in time when a data symbol is received at each antenna. There is great potential to get around the phase noise problem by design of smart transmission physical layer schemes and receiver algorithms.

**Internal power consumption:** Massive MIMO offers the potential to reduce the radiated power 1000 times and at the same time drastically scale up data rates. However, in practice, the total power consumed must be considered, including the cost of baseband signal processing. Much research must be invested in highly parallel, perhaps dedicated, hardware for the baseband signal processing.

**Channel characterization:** There are additional properties of the channel to consider when using massive MIMO instead of conventional MIMO. To facilitate a realistic performance assessment of massive MIMO systems, it is necessary to have channel models that reflect the true behavior of the radio channel (i.e., the propagation channel including effects of realistic antenna arrangements). It is also important to develop more sophisticated analytical channel models. Such models need not necessarily be correct in every fine detail, but they must capture the essential behavior of the channel. For example, in conventional MIMO the Kronecker model is widely used to model channel correlation. This model is not an exact representation of reality, but provides a useful model for certain types of analysis despite its limitations. A similar way of thinking could probably be adopted for massive MIMO channel modeling.

**Cost of reciprocity calibration:** TDD will require reciprocity calibration. How often must this be done, and what is the best way of doing it? What is the cost, in terms of time- and frequency resources needed to do the calibration, and in terms of additional hardware components needed?

**Pilot contamination:** It is likely that pilot contamination imposes much more severe limitations on massive MIMO than on traditional MIMO systems. We have discussed some of the issues in detail and outlined some of the most relevant research directions earlier.

**Non-CSf@TX operation:** Before a link has been established with a terminal, the base station has no way of knowing the channel response to the terminal. This means that no array beamforming gain can be harnessed. In this case, probably some form of space-time block coding...
is optimal. Once the terminal has been contact-
ed and sent a pilot, the base station can learn
the channel response and operate in coherent
MU-MIMO beamforming mode, reaping the
power gains offered by having a very large array.

New deployment scenarios: It is considered
extraordinarily difficult to introduce a radical
new wireless standard. One possibility is to intro-
duce dedicated applications of massive MIMO
technology that do not require backward com-
patibility. For example, as discussed earlier, in
rural areas, a billboard-sized array could provide
20Mb/s to each of 1000 homes using special equipment that would be used solely for
this application. Alternatively, a massive array
could provide the backhaul for base stations
that serve small cells in a densely populated area.

Thus, rather than thinking of massive MIMO as
a competitor to LTE, it can be an enabler for
something that was just never before considered
possible with wireless technology.

System studies and relation to small-cell and
eheterogeneous network solutions: The driving
motivation of massive MIMO is to simultaneously
and drastically increase data rates and overall
energy efficiency. Other potential ways of reach-
ing this goal are network densification by the
deployment of small cells, resulting in a hetero-
egeneous architecture, or coordination of the
transmission of multiple individual base stations.

From a purely fundamental perspective, the ulti-
mately limiting factor of the performance of any
wireless network appears to be the availability of
good enough channel state information (CSI) to
facilitate phase-coherent processing at multiple antennas or multiple access points [20]. Consider-
ing factors like mobility, Doppler shifts, phase
noise, and clock synchronization, acquiring high-
quality CSI seems to be easier with a collocated
massive array than in a system where the anten-
as are distributed over a large geographical
area. But at the same time, a distributed array
or small cell solution may offer substantial path loss
gains and would also provide some diversity
against shadow fading. The deployment costs of
a massive MIMO array and a distributed or
small cell system are also likely to be very differ-
ent. Hence, both communication-theoretic and
techno-economic studies are needed to conclu-
sively determine which approach is superior.

However, it is likely that the winning solution
will comprise a combination of all available tech-
nologies.

Prototype development: While massive
MIMO is in its infancy, basic prototyping work
on various aspects of the technology is going on
different parts of the world. The Argos testbed
[3] was developed at Rice University in coopera-
tion with Alcatel-Lucent, and shows the basic
feasibility of the massive MIMO concept using
64 coherently operating antennas. In particular,
the testbed shows that TDD operation relying on
channel reciprocity is possible. One of the
virtues of the Argos testbed in particular is that
it is entirely modular and scalable, and built
around commercially available hardware (the
WARP platform). Other test systems around the
world have also demonstrated the basic feasibil-
ity of scaling up the number of antennas. The
Ngara testbed in Australia [21] uses a 32-ele-
ment base station array to serve up to 18 users
simultaneously with true spatial multiplexing.

Continued testbed development is highly desired to
both prove the massive MIMO concept with
even larger numbers of antennas and discover
potentially new issues that urgently need research.

CONCLUSIONS AND OUTLOOK

In this article we have highlighted the large
potential of massive MIMO systems as a key
enabling technology for future beyond fourth
generation (4G) cellular systems. The technology
offers huge advantages in terms of energy effi-
ciency, spectral efficiency, robustness, and relia-
bility. It allows for the use of low-cost hardware
at both the base station and the mobile unit side.
At the base station the use of expensive and
powerful, but power-inefficient, hardware is
replaced by massive use of parallel low-cost low-
power units that operate coherently together.
There are still challenges ahead to realize the
full potential of the technology, for example,
computational complexity, realization of dis-
tributed processing algorithms, and synchroniza-
tion of the antenna units. This gives researchers
in both academia and industry a gold mine of
totally new research problems to tackle.

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There are still challenges ahead to realize the full potential of the technology, for example, when it comes to computational complexity, realization of distributed processing algorithms, and synchronization of the antenna units.